

**Specification for radio disturbance and immunity  
measuring apparatus and methods –**

**Part 1-4:**

**Radio disturbance and immunity measuring  
apparatus – Ancillary equipment –  
Radiated disturbances**

*This **English-language** version is derived from the original **bilingual** publication by leaving out all French-language pages. Missing page numbers correspond to the French-language pages.*

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**16-1-4**

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INTERNATIONAL SPECIAL COMMITTEE ON RADIO INTERFERENCE

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**Specification for radio disturbance and immunity  
measuring apparatus and methods –**

**Part 1-4:  
Radio disturbance and immunity measuring  
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Radiated disturbances**

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INTERNATIONAL ELECTROTECHNICAL COMMISSION  
INTERNATIONAL SPECIAL COMMITTEE ON RADIO INTERFERENCE

**SPECIFICATION FOR RADIO DISTURBANCE AND IMMUNITY  
MEASURING APPARATUS AND METHODS –**

**Part 1-4: Radio disturbance and immunity measuring apparatus –  
Ancillary equipment – Radiated disturbances**

FOREWORD

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International Standard CISPR 16-1-4 has been prepared by CISPR subcommittee A: Radio interference measurements and statistical methods.

This second edition of CISPR 16-1-4 cancels and replaces the first edition published in 2003, amendment 1 (2004) and amendment 2 (2005).

The document CISPR/A/710/FDIS, circulated to the National Committees as amendment 3, led to the publication of the new edition.

The text of this standard is based on the first edition, its Amendment 1, Amendment 2 and the following documents:

FDIS	Report on voting
CISPR/A/710/FDIS	CISPR/A/722/RVD

Full information on the voting for the approval of this standard can be found in the report on voting indicated in the above table.

This publication has been drafted in accordance with the ISO/IEC Directives, Part 2.

A list of all parts of CISPR 16 series, under the general title *Specification for radio disturbance and immunity measuring apparatus and methods*, can be found on the IEC website.

CISPR 16-1 consists of the following parts, under the general title *Specification for radio disturbance and immunity measuring apparatus and methods – Radio disturbance and immunity measuring apparatus*:

Part 1-1: Measuring apparatus

Part 1-2: Ancillary equipment – Conducted disturbances

Part 1-3: Ancillary equipment – Disturbance power

Part 1-4: Ancillary equipment – Radiated disturbances

Part 1-5: Antenna calibration test sites for 30 MHz to 1 000 MHz

The committee has decided that the contents of this publication will remain unchanged until the maintenance result date indicated on the IEC web site under "<http://webstore.iec.ch>" in the data related to the specific publication. At this date, the publication will be

- reconfirmed,
- withdrawn,
- replaced by a revised edition, or
- amended.

## **SPECIFICATION FOR RADIO DISTURBANCE AND IMMUNITY MEASURING APPARATUS AND METHODS –**

### **Part 1-4: Radio disturbance and immunity measuring apparatus – Ancillary equipment – Radiated disturbances**

#### **1 Scope**

This part of CISPR 16 is designated a basic standard, which specifies the characteristics and performance of equipment for the measurement of radiated disturbances in the frequency range 9 kHz to 18 GHz.

Specifications for ancillary apparatus are included for: antennas and test sites, TEM cells, and reverberating chambers.

The requirements of this publication must be complied with at all frequencies and for all levels of radiated disturbances within the CISPR indicating range of the measuring equipment.

Methods of measurement are covered in Part 2-3, and further information on radio disturbance is given in Part 3 of CISPR 16. Uncertainties, statistics and limit modelling are covered in Part 4 of CISPR 16.

#### **2 Normative references**

The following referenced documents are indispensable for the application of this document. For dated references, only the edition cited applies. For undated references, the latest edition of the referenced document (including any amendments) applies.

CISPR 16-1-1, *Specification for radio disturbance and immunity measuring apparatus and methods – Part 1-1: Radio disturbance and immunity measuring apparatus – Measuring apparatus*

CISPR 16-2-3, *Specification for radio disturbance and immunity measuring apparatus and methods – Part 2-3: Methods of measurement of disturbances and immunity – Radiated disturbance measurements*

CISPR 16-3, *Specification for radio disturbance and immunity measuring apparatus and methods – Part 3: CISPR technical reports*

CISPR 16-4 (all parts), *Specification for radio disturbance and immunity measuring apparatus and methods – Uncertainties, statistics and limit modelling*

CISPR 16-4-2:2003, *Specification for radio disturbance and immunity measuring apparatus and methods – Part 4-2: Uncertainties, statistics and limit modelling – Uncertainty in EMC measurements*

IEC 60050-161, *International Electrotechnical Vocabulary (IEV) – Chapter 161: Electromagnetic compatibility*

### 3 Terms and definitions

For the purposes of this document, the following terms and definitions apply. Also see IEC 60050(161).

#### 3.1

##### **bandwidth**

##### **$B_n$**

width of the overall selectivity curve of the receiver between two points at a stated attenuation, below the midband response

NOTE The bandwidth is represented by the symbol  $B_n$ , where  $n$  is the stated attenuation in decibels.

#### 3.2

##### **CISPR indicating range**

range specified by the manufacturer which gives the maximum and the minimum meter indications within which the receiver meets the requirements of this part of CISPR 16

#### 3.3

##### **calibration test site**

##### **CALTS**

open area test site with metallic ground plane and tightly specified site attenuation performance in horizontal and vertical electric field polarization

NOTE 1 A CALTS is used for determining the free-space antenna factor of an antenna.

NOTE 2 Site attenuation measurements of a CALTS are used for comparison to corresponding site attenuation measurements of a compliance test site, in order to evaluate the performance of the compliance test site.

#### 3.4

##### **compliance test site**

##### **COMTS**

environment which assures valid, repeatable measurement results of disturbance field strength from equipment under test for comparison to a compliance limit

#### 3.5

##### **antenna**

that part of a transmitting or receiving system that is designed to radiate or to receive electromagnetic waves in a specified way

NOTE 1 In the context of this standard, the balun is a part of the antenna.

NOTE 2 See also the term "wire antenna".

#### 3.6

##### **balun**

passive electrical network for the transformation from a balanced to an unbalanced transmission line or device or vice versa

#### 3.7

##### **free-space-resonant dipole**

wire antenna consisting of two straight colinear conductors of equal length, placed end to end, separated by a small gap, with each conductor approximately a quarter-wavelength long such that at the specified frequency the input impedance of the wire antenna measured across the gap is pure real when the dipole is located in the free space

NOTE 1 In the context of this standard, this wire antenna connected to the balun is also called the "test antenna".

NOTE 2 This wire antenna is also referred to as "tuned dipole".

### 3.8

#### **site attenuation**

insertion loss determined by a two-port measurement, when a direct electrical connection between the generator output and receiver input is replaced by transmitting and receiving antennas placed at the specified positions

### 3.9

#### **test antenna**

combination of the free-space-resonant dipole and the specified balun

NOTE For the purpose of this standard only.

### 3.10

#### **wire antenna**

a specified structure consisting of one or more metallic wires or rods for radiating or receiving electromagnetic waves

NOTE A wire antenna does not contain a balun.

### 3.11

#### **fully anechoic room**

##### **FAR**

shielded enclosure, the internal surfaces of which are lined with radio-frequency absorbing material (i.e. RF absorber), which absorbs electromagnetic energy in the frequency range of interest

### 3.12

#### **quasi-free space test-site**

test-site for which the site attenuation measured with vertically polarized tuned dipoles deviates by no more than  $\pm 1$  dB from the calculated free-space attenuation at any frequency

### 3.13

#### **test volume**

volume in the FAR in which the EUT is positioned

NOTE In this volume the quasi-free space condition is met and this volume is typically 0,5 m or more from the absorbing material of the FAR.

## 4 Antennas for measurement of radiated radio disturbance

The antenna and the circuits inserted between it and the measuring receiver shall not appreciably affect the overall characteristics of the measuring receiver. When the antenna is connected to the measuring receiver, the measuring system shall comply with the bandwidth requirements of CISPR 16-1-1 appropriate to the frequency band concerned.

The antenna shall be substantially plane polarized. It shall be orientable so that all polarizations of incident radiation can be measured. The height of the centre of the antenna above ground may have to be adjustable according to a specific test procedure.

For additional information about the parameters of broadband antennas see Annex A.

### 4.1 Accuracy of field-strength measurements

The accuracy of field-strength measurement of a uniform field of a sine-wave shall be better than  $\pm 3$  dB when an antenna meeting the requirements of this subclause is used with a measuring receiver meeting the requirements of CISPR 16-1-1.

NOTE This requirement does not include the effect due to a test site.

## 4.2 Frequency range 9 kHz to 150 kHz

Experience has shown that, in this frequency range, it is the magnetic field component that is primarily responsible for observed instances of interference.

### 4.2.1 Magnetic antenna

For measurement of the magnetic component of the radiation, either an electrically-screened loop antenna of dimension such that the antenna can be completely enclosed by a square having sides of 60 cm in length, or an appropriate ferrite-rod antenna, may be used.

The unit of the magnetic field strength is  $\mu\text{A/m}$  or, in logarithmic units,  $20 \log(\mu\text{A/m}) = \text{dB}(\mu\text{A/m})$ . The associated emission limit shall be expressed in the same units.

NOTE Direct measurements can be made of the strength of the magnetic component, in  $\text{dB}(\mu\text{A/m})$  or  $\mu\text{A/m}$  of a radiated field under all conditions, that is, both in the near field and in the far field. However, many field strength measuring receivers are calibrated in terms of the equivalent plane wave electric field strength in  $\text{dB}(\mu\text{V/m})$ , i.e. assuming that the ratio of the  $E$  and  $H$  components is  $120 \pi$  or  $377 \Omega$ . This assumption is justified under far-field conditions at distances from the source exceeding one sixth of a wavelength ( $\lambda/2\pi$ ), and in such cases the correct value for the  $H$  component can be obtained by dividing the  $E$  value indicated on the receiver by 377, or by subtracting 51,5 dB from the  $E$  level in  $\text{dB}(\mu\text{V/m})$  to give the  $H$  level in  $\text{dB}(\mu\text{A/m})$ .

It should be clearly understood that the above fixed  $E$  and  $H$  ratio applies only under far-field conditions.

To obtain the reading of  $H$  ( $\mu\text{A/m}$ ), the reading  $E$  ( $\mu\text{V/m}$ ) is divided by  $377 \Omega$ :

$$H (\mu\text{A/m}) = E (\mu\text{V/m}) \times 377 \Omega \quad (1)$$

To obtain the reading of  $H$   $\text{dB}(\mu\text{A/m})$ , 51,5  $\text{dB}(\Omega)$  is subtracted from the reading  $E$   $\text{dB}(\mu\text{V/m})$ :

$$H \text{ dB}(\mu\text{A/m}) = E \text{ dB}(\mu\text{V/m}) - 51,5 \text{ dB}(\Omega) \quad (2)$$

The impedance  $Z = 377 \Omega$ , with  $20 \log Z = 51,5 \text{ dB}(\Omega)$ , used in the above conversions is a constant originating from the calibration of field strength measuring equipment indicating the magnetic field in  $\mu\text{V/m}$  (or  $\text{dB}(\mu\text{V/m})$ ).

### 4.2.2 Balance of antenna

The balance of the antenna shall be such that, when the antenna is rotated in a uniform field, the level in the cross-polarization direction is at least 20 dB below that in the parallel polarization direction.

## 4.3 Frequency range 150 kHz to 30 MHz

### 4.3.1 Electric antenna

For the measurement of the electric component of the radiation, either a balanced or an unbalanced antenna may be used. If an unbalanced antenna is used, the measurement will refer only to the effect of the electric field on a vertical rod antenna. The type of antenna used shall be stated with the results of the measurements.

Information pertaining to calculating the performance characteristics of a 1 m length monopole (rod) antenna and the characterization of its matching network is specified in Annex B.

Where the distance between the source of radiation and the antenna is 10 m or less, the total length of the antenna shall be 1 m. For distances greater than 10 m the preferred antenna length is 1 m, but in no case shall it exceed 10 % of the distance.

The unit of electric field strength shall be  $\mu\text{V/m}$  or, in logarithmic units,  $20 \log(\mu\text{V/m}) = \text{dB}(\mu\text{V/m})$ . The associated emission limit shall be expressed in the same units.

### 4.3.2 Magnetic antenna

For the measurement of the magnetic component of the radiation, an electrically-screened loop antenna, as described in 4.2.1 shall be used.

Tuned electrically balanced loop antennas may be used to make measurements at lower field strengths than untuned electrically-screened loop antennas.

### 4.3.3 Balance of antenna

If a balanced electric or a magnetic antenna is used, it shall comply with the requirement of 4.2.2.

## 4.4 Frequency range 30 MHz to 300 MHz

### 4.4.1 Electric antenna

The reference antenna shall be a balanced dipole.

#### 4.4.1.1 Balanced dipole

For frequencies 80 MHz or above, the antenna shall be resonant in length, and for frequencies below 80 MHz it shall have a length equal to the 80 MHz resonant length and shall be tuned and matched to the feeder by a suitable transforming device. Connection to the input of the measuring apparatus shall be made through a symmetric-asymmetric transformer arrangement.

#### 4.4.1.2 Shortened dipole

A dipole shorter than a half wavelength may be used provided:

- a) the total length is greater than 1/10 of a wavelength at the frequency of measurement;
- b) it is connected to a cable sufficiently well matched at the receiver end to ensure a voltage standing wave ratio (v.s.w.r.) on the cable of less than 2.0 to 1. The calibration shall take account of the v.s.w.r.;
- c) it has a polarization discrimination equivalent to that of a tuned dipole (see 4.4.2). To obtain this, a balun may be helpful;
- d) for determination of the measured field strength, a calibration curve (antenna factor) is determined and used in the measuring distance (i.e., at a distance of at least three times the length of the dipole);

NOTE The antenna factors thus obtained should make it possible to fulfil the requirement of measuring uniform sine-wave fields with an accuracy not worse than  $\pm 3$  dB. Examples of calibration curves are given in Figure 1 which shows the theoretical relation between field strength and receiver input voltage for a receiver of input impedance of  $50 \Omega$ , and for various  $l/d$  ratios. On these figures, the balun is considered as an ideal 1:1 transformer. It should be noted, however, that these curves do not account for the losses of the balun, the cable and any mismatch between the cable and the receiver.

- e) in spite of the sensitivity loss of the field-strength meter due to a high antenna factor attributed to the shortened length of the dipole, the measuring limit of the field-strength meter (determined for example by the noise of the receiver and the transmission factor of the dipole) shall remain at least 10 dB below the level of the measured signal.

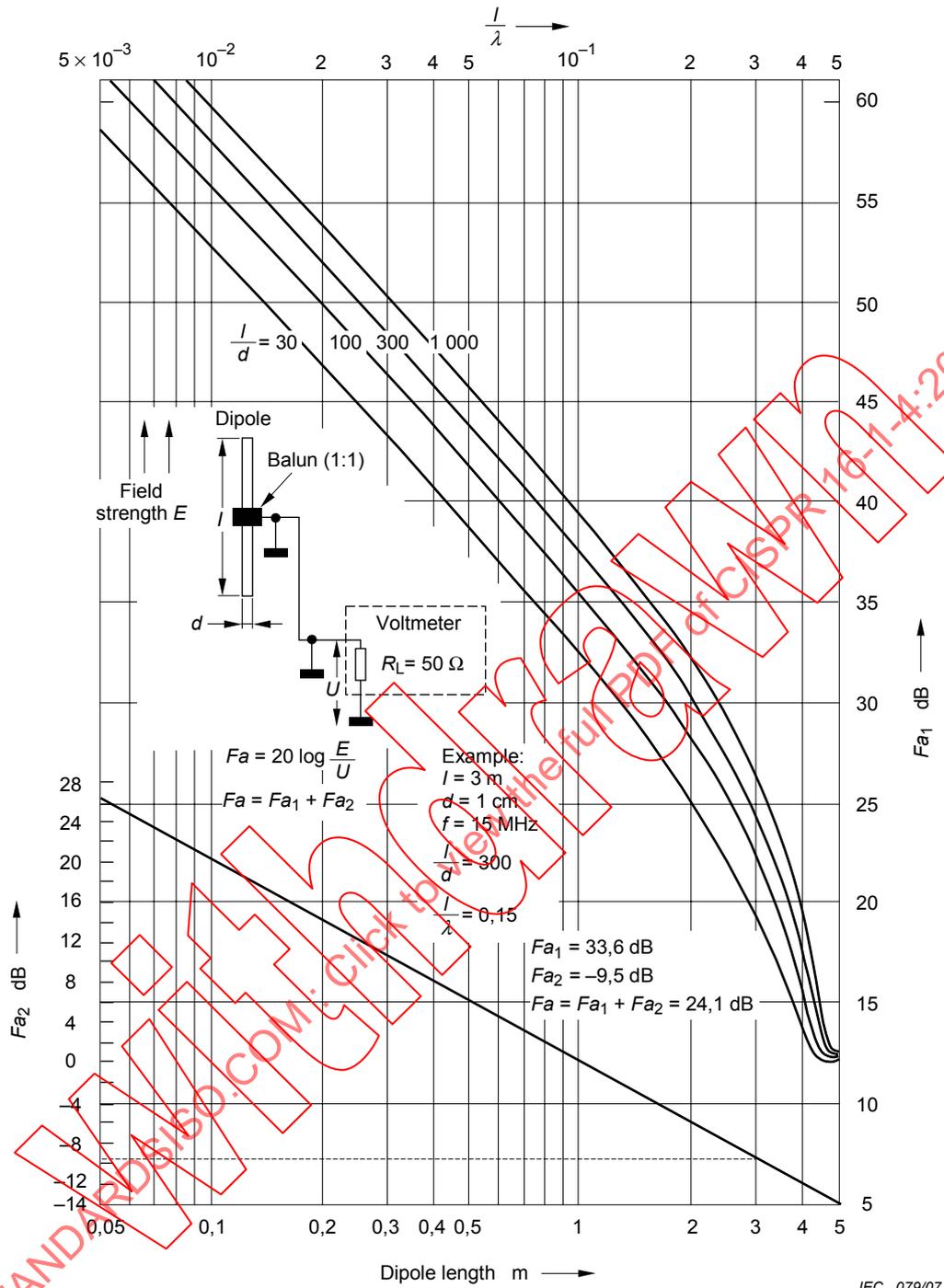


Figure 1 – Short dipole antenna factors for  $R_L = 50 \Omega$

#### 4.4.1.3 Broadband antenna

A broadband antenna may be used, provided that it meets the requirements given in 4.5.2 for a complex antenna.

#### 4.4.2 Balance of antenna

##### 4.4.2.1 Introduction

In radiated emission measurements, common-mode (CM) currents may be present on the cable attached to the receiving antenna (the antenna cable). In turn, these CM currents create EM fields which may be picked up by the receiving antenna. Consequently, the radiated emission measuring results may be influenced.

The major contributions to the antenna cable CM currents stem from

- a) the electric field generated by the EUT, if that field has a component parallel to the antenna cable, and
- b) the conversion of the differential mode (DM) antenna signal (the desired signal) into a CM signal by the imperfection of the balun of the receiving antenna.

This subclause considers the balun contribution. Contribution a) is under consideration (see last sentence of Note 1 of 4.4.2.2).

In general, log-periodic dipole array antennas do not exhibit significant DM/CM conversion and the following check applies to dipoles, biconical antennas and bicone/log hybrid antennas.

##### 4.4.2.2 Balun DM/CM conversion check

The following method describes the measurement of two voltages,  $U_1$  and  $U_2$ , in the frequency range for which the receiving antenna is to be used. The ratio of these voltages, both expressed in identical units (e.g., dBuV), is a measure for the DM/CM conversion.

- 1) Set the receiving antenna under test vertically polarized with its centre at a height of 1,5 m above the ground plane. Lay the cable horizontally for  $1,5 \text{ m} \pm 0,1 \text{ m}$  behind the rear active element of the antenna and then drop it vertically by a height of at least 1,5 m to the ground plane.
- 2) Place a second (transmitting) antenna vertically polarized at a horizontal distance of 10 m from the centre of the antenna under test with its tip 0,10 m from the ground plane. If the range of the site used for emission testing is 3 m, do this check using a distance of 3 m (if the conversion check has already been made at 10 m distance and shows a change of less than  $\pm 0,5 \text{ dB}$ , it is not necessary to take a separate measurement at 3 m). The specification of the transmitting antenna shall include the frequency range of the antenna under test.
- 3) Connect the transmitting antenna to a signal source, for example, a tracking generator, set the level of that generator in such a way that, over the frequency range of interest, the signal-to-ambient noise at the receiver is larger than 10 dB.
- 4) Record the voltage  $U_1$  at the receiver over the frequency range of interest.
- 5) Invert the receiving antenna (rotate that antenna through  $180^\circ$ ) without changing anything else in the set-up, in particular the receiving antenna cable, and without changing the setting of the signal source.
- 6) Record the voltage  $U_2$  at the receiver over the frequency range.

7) The DM/CM conversion is sufficiently low if  $|20 \log (U_1/U_2)| < 1$  dB.

NOTE 1 If the DM/CM conversion criterion is not met, ferrite rings around the antenna cable may reduce the DM/CM conversion. The addition of ferrites on the antenna cable may also be used to verify whether contribution a) has a non-negligible effect. Repeat the test with four ferrites spaced approximately 20 cm apart. If the criterion is met by using these rings, they should be present in the actual emission measurement. Likewise, the interaction with the cable can be reduced by extending the cable several metres behind the antenna before dropping to ground.

NOTE 2 If the receiving antenna is to be used in a fully anechoic chamber, the DM/CM check may be performed in that room with the receiving antenna at its usual location and the transmitting antenna in the centre of the test volume of that room. The room must comply with the  $\pm 4$  dB criterion.

NOTE 3 The measuring site of which the ground plane forms a part, or the fully anechoic room, should comply with their respective NSA (normalized site attenuation) requirements.

NOTE 4 The horizontal distance of 1,5 m over which the antenna cable runs horizontally behind the centre of the antenna should be kept as a minimum during actual vertically polarized radiated emissions measurements.

NOTE 5 It is not necessary to define a test set-up strictly because this effect is in large part due to the interaction of the antenna and the part of input cable that lies parallel to the antenna elements. There is a much smaller effect which is dependent on the uniformity of the field incident on the antenna in normal EMC set-ups on an OATS or in a fully anechoic room.

NOTE 6 For baluns which have the receive cable connector mounted on the side (90° to the antenna boom), a right angle connector should be used to reduce the movement of the cable.

#### 4.4.3 Cross-polar performance of antenna

When an antenna is placed in a plane-polarized electromagnetic field, the terminal voltage when the antenna and field are cross-polarized shall be at least 20 dB below the terminal voltage when they are co-polarized. It is intended that this test apply to log-periodic dipole array (LPDA) antennas for which the two halves of each dipole are in eschelon. The majority of testing with such antennas is above 200 MHz, but the requirement applies below 200 MHz. This test is not intended for in-line dipole and biconical antennas because a cross-polar rejection greater than 20 dB is intrinsic to their symmetrical design. Such antennas and horn antennas must have a cross-polar rejection greater than 20 dB and a type test by the manufacturer should confirm this.

In order to achieve quasi-free space conditions, a high-quality anechoic chamber or towers of sufficient height above ground on an outdoor range can be used. To minimize ground reflections, set the antennas vertically polarized. A plane wave shall be set up at the antenna under test. The separation between the centre of the antenna under test and the source antenna shall be greater than one wavelength.

NOTE A good-quality site is needed to set up a plane wave at the antenna under test. The cross-polar discrimination afforded by the plane wave can be proven by transmitting between a pair of horn antennas or open-ended waveguides and checking that the combination of site error and inherent cross-polar performance of one horn antenna yields a suppression of the horizontal component by more than 30 dB. If the site errors are very low and if the horn antennas have identical performance, the cross-polar performance of one horn is approximately 6 dB lower than the combined cross-polar coupling of the pair of horns.

An interfering signal 20 dB lower in level than the desired signal gives a maximum error on the desired signal of  $\pm 0,9$  dB. The maximum error occurs when the cross-polar signal is in phase with the co-polar signal. If the cross-polar response of the LPDA is worse than 20 dB, the operator must calculate the uncertainty and declare it with the result. For example a cross-polar level of 14 dB implies a maximum uncertainty of +1,6 dB to –1,9 dB. Take the larger value and assume a U-shaped distribution when calculating the standard uncertainty.

To add a signal of 0 dB to another of –14 dB, first convert to relative voltages by dividing by 20 and taking the anti-log. Then add the smaller signal to the unity signal. Take the log and multiply by 20. The result is the positive decibel error. Repeat, but subtracting the smaller signal from the unity signal to give the negative decibel error.

For the purpose of calculating the uncertainty of the result of a radiated emission, if the signal level measured in one polarization exceeds the signal measured in the orthogonal polarization by 6 dB or more, then an LPDA whose cross-polar discrimination is only 14 dB will have been deemed to have met the specification of 20 dB. If the difference between the VP and HP signal levels is less than 6 dB, additional uncertainty must be calculated if the sum of this difference and the cross-polarization is less than 20 dB.

#### 4.5 Frequency range 300 MHz to 1 000 MHz

##### 4.5.1 Electric antenna

If a dipole antenna is used, it shall meet the requirements of 4.4.1.1 and 4.4.2.

##### 4.5.2 Complex antenna

Since, at the frequencies in the range 300 MHz to 1 000 MHz, the sensitivity of the simple dipole antenna is low, a more complex antenna may be used. Such antenna shall be as follows.

- a) The antenna shall be substantially plane polarized. This shall be checked in the same manner as for the balance of a simple dipole antenna.
- b) The main lobe of the radiation pattern of the antenna shall be such that the response in the direction of the direct ray and that in the direction of the ray reflected from the ground do not differ by more than 1 dB.

To ensure this condition, the total vertical angular aperture  $2\varphi$  of the measuring antenna, within which the antenna gain is within 1 dB of its maximum, shall be such that:

- 1) if the measuring antenna is maintained in a horizontally direct position:

$$\varphi > \tan^{-1} [(h_1 + h_2)/d]$$

- 2) if the measuring antenna is tilted towards earth in the optimum position (so that direct and reflected rays are included within the aperture  $2\varphi$ ):

$$2\varphi > \tan^{-1} [(h_1 + h_2)/d] - \tan^{-1} [(h_1 - h_2)/d]$$

where

$h_1$  is the measuring antenna height;

$h_2$  is the height of the device under test;

$d$  is the horizontal distance between the measuring antenna and the device under test.

The pattern of the antenna shall be checked in the horizontal plane while orienting it for vertical polarization. It shall be assumed that the pattern and, in particular, the angular aperture  $2\varphi$  is the same when horizontally polarized as when measured with the vertical polarization.

It is essential that the variation of the effective distance of the antenna from the source and its gain with frequency be taken into account.

- c) The voltage standing-wave ratio of the antenna with the antenna feeder connected and measured from the receiver end shall not exceed 2,0 to 1.
- d) A calibration factor shall be given making it possible to fulfil the requirements of 4.1.

#### 4.6 Frequency range 1 GHz to 18 GHz

Radiated emissions measurements above 1 GHz shall be made using calibrated, linearly polarized antennas. These include double-ridged guide horns, rectangular wave guide horns, pyramidal horns, optimum gain horns and standard gain horns. The "beam" or main lobe of the pattern of any antenna used shall be large enough to encompass the EUT when located at the measuring distance, or provisions shall be made for "scanning" the EUT to locate the direction or source of its radiated emissions. The width of the main lobe is defined as the 3 dB beamwidth of the antenna, and information enabling the determination of this parameter should be given in the antenna documentation. The aperture dimensions of these horn antennas shall be small enough so that the measurement distance  $R_m$  in metres is equal to or greater than the following minimum distance:

$$R_m \geq D^2/2\lambda$$

where

$D$  is the largest dimension of the aperture in metres of the antenna;

$\lambda$  is the free space wavelength in metres at the frequency of measurement.

In case of dispute, measurements made with a standard gain horn antenna or a similar precisely calibrated horn antenna shall take precedence.

NOTE Any calibrated, linearly polarized antenna, e.g. a log periodic dipole array, may be used to make these measurements. The gain of many antennas other than horn antennas in this frequency range may be inadequate if the antennas are used with spectrum analyzers or older radio noise meters. The tester should assure that the overall measurement sensitivity is at least 6 dB below the applicable limit at the measurement distance in use, and that any means used to improve sensitivity, e.g. a preamplifier, does not cause distortion, spurious signals, or other overload problems. Since a log periodic dipole array has a much wider beamwidth than a horn antenna, reflections from the ground plane may cause significant error in measurements that are made with a log periodic dipole array.

#### 4.7 Special antenna arrangements

##### 4.7.1 Loop antenna system

In the frequency range 9 kHz to 30 MHz, the interference capability of the magnetic field component of the radiation of a single (EUT) can be determined by using a special loop antenna system (LAS). In the LAS, this capability is measured in terms of the currents induced by the magnetic field in the loop antennas of the LAS. The LAS allows indoor measurements.

The LAS consists of three circular, mutually perpendicular large-loop antennas (LLAs), having a diameter of 2 m, supported by a non-metallic base. A full description of the LAS is given in Annex C.

The EUT is positioned in the centre of the LAS. The maximum dimensions of the EUT are limited so that the distance between the EUT and an LLA is at least 0,20 m. Guidelines for the routing of signal cables are given in Clause C.3, Note 2, and Figure C.6. Cables should be routed together and leave the loop volume in the same octant of the cell and no closer than 0,4 m to any of the LAS loops.

The three mutually perpendicular LLAs allow measurement of the interference capability of all polarizations of the radiated field with the prescribed accuracy, and without rotation of the EUT or changing the orientation of the LLAs.

Each of the three LLAs shall comply with the validation requirements given in Clause C.5.

NOTE Circular LLAs having a diameter different from the standardized diameter of 2 m may be used, provided their diameter  $D \leq 4$  m and the distance between the EUT and a LA is at least  $0,10(D)$  m. Correction factors for non-standardized diameters are given in Clause C.6.

## 5 Test sites for measurement of radio disturbance field strength for the frequency range of 30 MHz to 1 000 MHz

An environment is required which assures valid, repeatable measurement results of disturbance field strength from equipment. For equipment which can only be tested in its place of use, different provisions have to be utilized.

### 5.1 Open area test site

Disturbance field-strength measurements are normally performed at an open area test site. Open area test sites are areas characteristic of cleared level terrain. Such test sites shall be void of buildings, electric lines, fences, trees, etc. and free from underground cables, pipelines, etc., except as required to supply and operate the equipment under test (EUT). Refer to Annex D for specific construction recommendations for open area test sites for electromagnetic field tests in the range of 30 MHz to 1 GHz. The site validation procedure for open area test sites is given in 5.6 with further details in Annex E. Annex F contains the acceptability criterion.

### 5.2 Weather protection enclosure

Weather protection is desirable if the test site is used throughout the year. A weather protection structure could either protect the whole test site including EUT and field strength measuring antenna or the EUT only. The materials used shall be RF transparent in order to cause no undesirable reflections and attenuation of the emitted field from the EUT.

The structure shall be shaped to allow easy removal of snow, ice or water. For further details, see Annex D.

### 5.3 Obstruction-free area

For open area test sites, an obstruction-free area surrounding the EUT and field-strength measuring antenna is required. The obstruction-free area should be free from significant scatterers of electromagnetic fields, and should be large enough so that scatterers outside the obstruction-free area will have little effect on the fields measured by the field-strength measuring antenna. To determine the adequacy of this area, site validation tests should be performed.

Since the magnitude of the field scattered from an object depends on many factors (size of the object, distance from the EUT, orientation with respect to the EUT, conductivity and permittivity of the object, frequency, etc.), it is impractical to specify a reasonable obstruction-free area which is necessary and sufficient for all applications. The size and shape of the obstruction-free area are dependent upon the measurement distance and whether or not the EUT will be rotated. If the site is equipped with a turntable, the recommended obstruction-free area is an ellipse with the receiving antenna and EUT at the two foci and having a major axis equal to twice the measurement distance and a minor axis equal to the product of the measurement distance and the square root of 3 (see Figure 2).

For this ellipse, the path of the undesired ray reflected from any object on the perimeter is twice the length of the direct ray path between the foci. If a large EUT is installed on the turntable, the obstruction-free area must be expanded so that the obstruction clearance distances exist from the perimeter of the EUT.

If the site is not equipped with a turntable, that is, the EUT is stationary, the recommended obstruction-free area is a circular area such that the radial distance from the boundary of the EUT to the boundary of the area is equal to the measurement distance multiplied by 1,5 (see Figure 3). In this case, the antenna is moved around the EUT at the separation distance.

The terrain within the obstruction-free area should be flat. Small slopes needed for adequate drainage are acceptable. The flatness of the metallic ground plane, if used, is discussed in Clause D.2. Measuring apparatus and test personnel should be situated outside the obstruction free area.

#### 5.4 Ambient radio frequency environment of a test site

The ambient radio frequency levels at a test site shall be sufficiently low compared to the levels of measurements to be performed. The quality of the site in this respect may be assessed in four categories, listed below in their order of merit:

- a) the ambient emissions are 6 dB or more below the measurement levels;
- b) some ambient emissions are within 6 dB of the measurement levels;
- c) some ambient emissions are above the measurement levels, but are either aperiodic (i.e., sufficiently long in time between transmissions to allow a measurement to be made) or continuous, but only on limited identifiable frequencies;
- d) the ambient levels are above the measurement levels over a large portion of the measurement frequency range and occurring continuously.

The selection of a test site should ensure that the accuracy of the measurement is maintained given the environment and the degree of engineering skill available.

NOTE For perfect results, an ambient level 20 dB below the emission level measured is recommended.

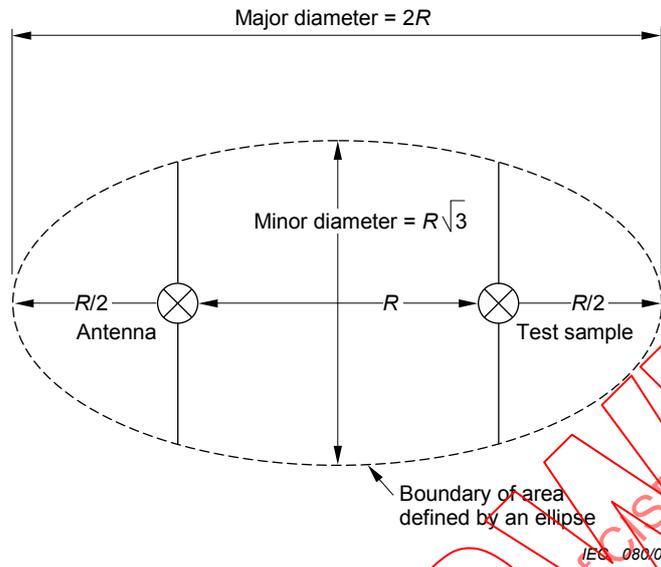


Figure 2 – Obstruction-free area of a test site with a turntable (see 5.3)

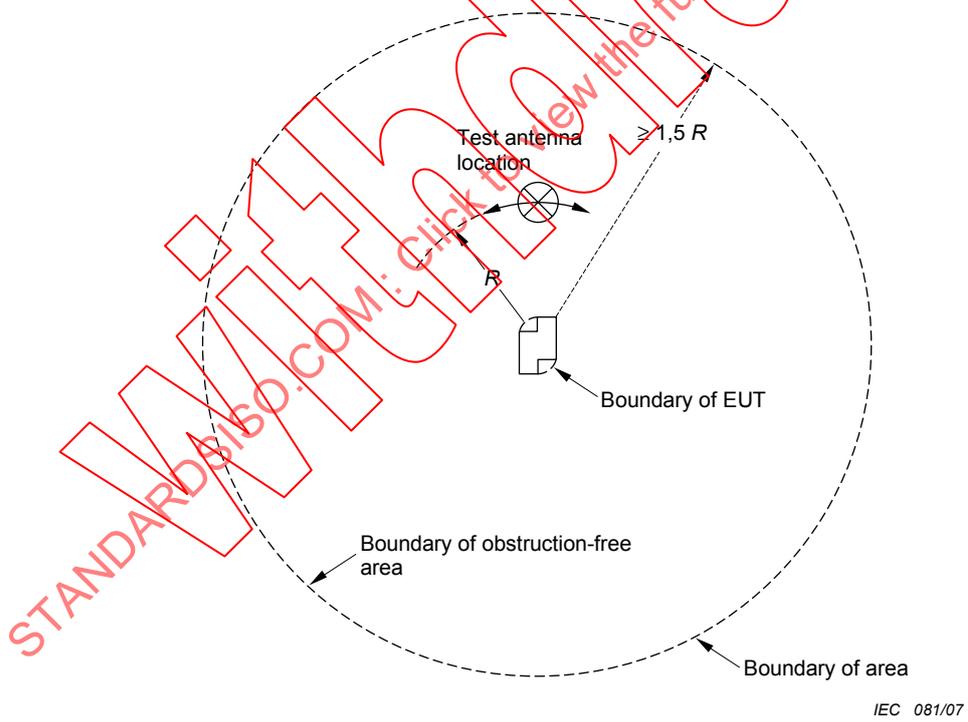


Figure 3 – Obstruction-free area with stationary EUT (see 5.3)

## 5.5 Ground plane

The ground plane may be composed of a wide range of material from earth to highly conductive, metallic material. The plane can be at earth level or elevated on a suitably sized platform or roof site. A metal ground plane is preferred, but for certain equipment and applications, it may not be recommended by certain product publications. Adequacy of the metal ground plane will be dependent on whether the test site meets the site validation requirements in 5.6. If no metallic material is used, caution is required to select a site that does not change its reflective characteristics with time, weather condition, or, due to buried metallic material such as pipes, conduits, and non-homogeneous soil. Such sites generally give different site attenuation characteristics compared to those with metallic surfaces.

## 5.6 Open area site validation procedure

The validation procedure and the requirements for the normalized site attenuation given here are used to qualify a test site when a metallic ground plane is specified. For other test sites, the validation procedure is of an informative nature, and will in general also identify possible site irregularities that should be investigated. Validation procedures applicable to absorber lined rooms are given in subclause 5.8.

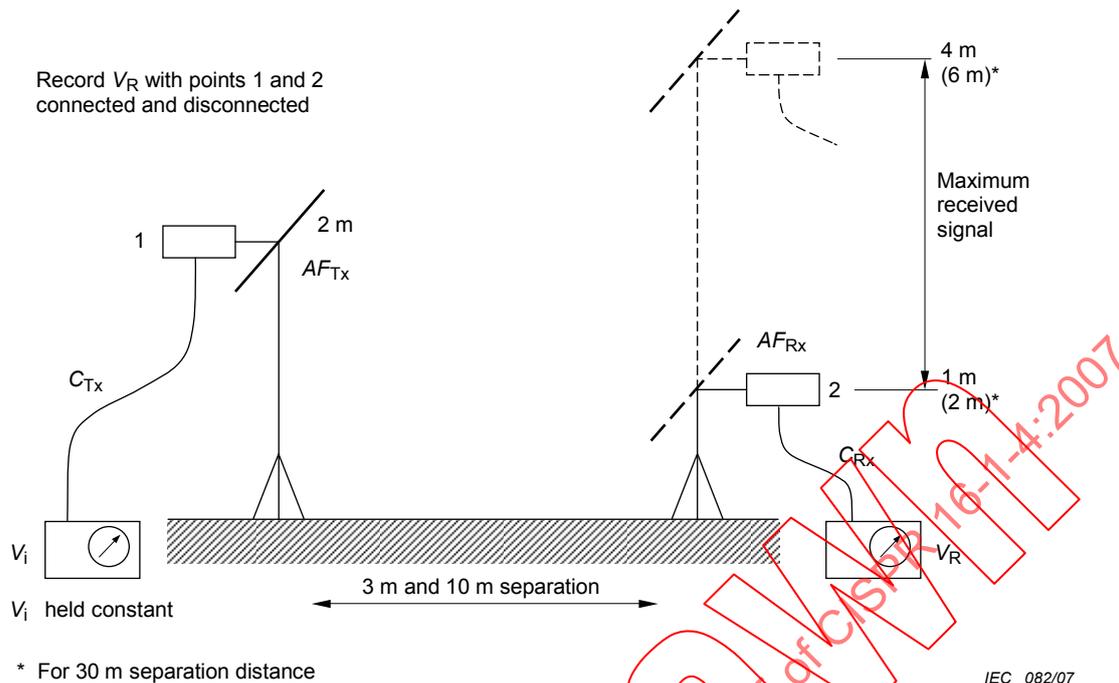
The validation of an open area test site is performed with two antennas oriented horizontally and vertically with respect to the ground, as shown in Figures 4 and 5, respectively. The open area site attenuation is obtained from the ratio of the source voltage ( $V_i$ ) connected to a transmitting antenna and the received voltage ( $V_r$ ) as measured on the receiving antenna terminals. The voltage measurements are performed in a  $50 \Omega$  system. Suitable corrections for cable losses is required if  $V_r$  and  $V_i$  are not measured at the input and output of the transmit and receive antenna, respectively. This site attenuation ratio is then divided by the product of the antenna factors for the two antennas used. The resulting answer is the normalized site attenuation (*NSA*) and is expressed in dB. The site is considered suitable when the measured vertical and horizontal *NSA*'s are within  $\pm 4$  dB of the values given in Tables E.1, E.2, and E.3, as appropriate. If the  $\pm 4$  dB criterion is exceeded, the test site must be investigated per Clause E.4.

NOTE The basis for the 4 dB site acceptability criterion is given in Annex F.

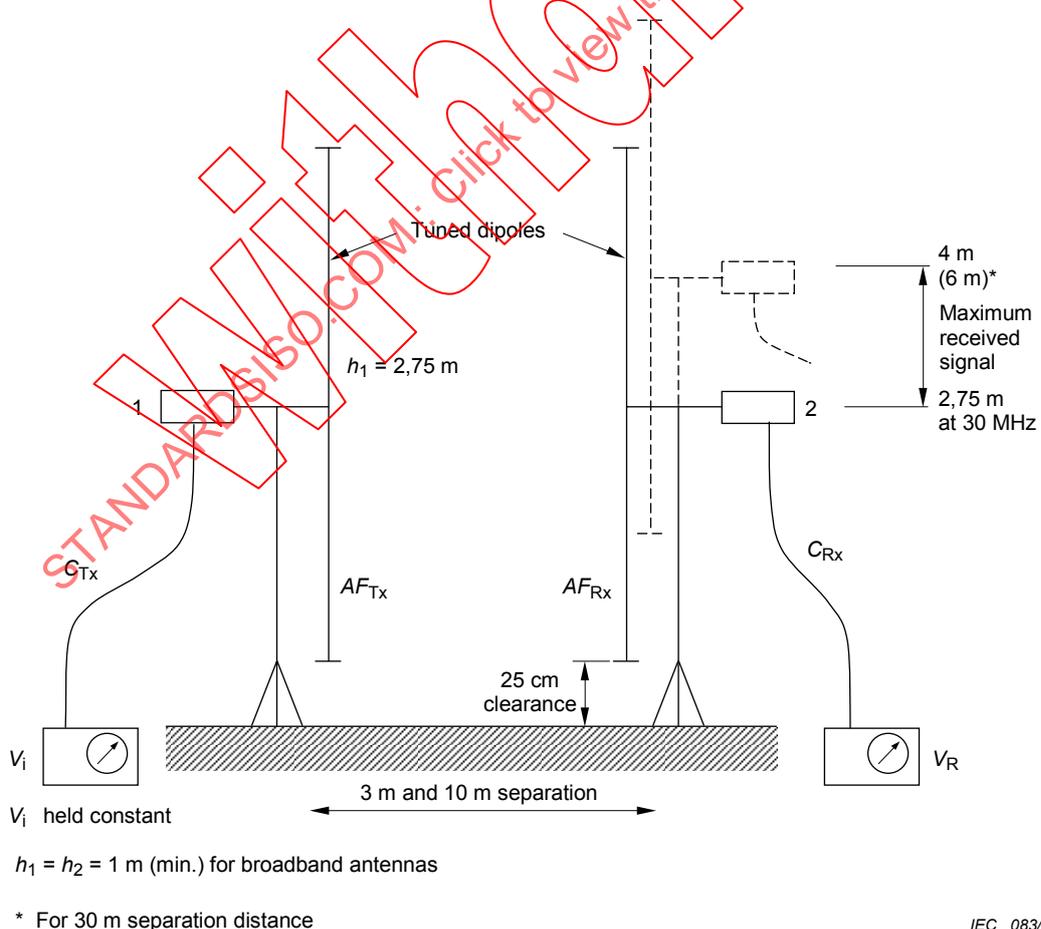
The deviation between a measured *NSA* value and the theoretical value shall not be used as a correction for a measured EUT field strength. This procedure shall be used only for validating a test site.

Table E.1 is used for broadband antennas such as biconical and log periodic arrays both horizontally and vertically aligned with respect to the ground plane. Table E.2 is for tuned half-wave dipoles aligned horizontally with respect to the ground plane. Table E.3 is for tuned half-wave dipoles vertically aligned with respect to the ground plane. Note that in Table E.3, there are restrictions in the scan height  $h_2$ . This takes into account the fact that the lowest tip of the receive dipole is kept 25 cm or more from the ground plane.

NOTE The reason for the different Tables E.1 and E.2/E.3 is that different geometrical parameters are chosen for a broadband antenna and a tuned half-wave dipole, primarily because of practical restrictions needed for the latter.



**Figure 4 – Configuration of equipment for measuring site attenuation in horizontal polarization (see 5.6 and Annex E)**



**Figure 5 – Configuration of equipment for measuring site attenuation in vertical polarization using tuned dipoles (see 5.6 and Annex E)**

*NSA* for frequencies other than those shown in the tables may be found using straight-line interpolation between the tabulated values.

The legend for each table is as follows:

- $R$  Horizontal separation distance between the projection of the transmit and receive antennas on the ground plane (metres).
- $h_1$  Height of the centre of the transmit antenna above the ground plane (metres).
- $h_2$  Range of heights of the centre of the receive antenna above the ground plane (metres). The maximum received signal in this height scan range is used for *NSA* measurements.
- $f_m$  Frequency in MHz.
- $A_N$  *NSA* (see Equation (1), below).

NOTE The spacing  $R$  between log-periodic *array* antennas is measured from the projection on to the ground plane of the mid-point of the longitudinal axis of each antenna.

It is recommended that horizontal *NSA* measurements be performed first. Since such measurements are less sensitive than that for vertical polarization in finding test anomalies, the measured *NSA* should readily be within  $\pm 4$  dB of that shown in Tables E.1, E.2 and E.3. If not, recheck measurement technique, instrumentation drift and antenna factor calibrations. If the  $\pm 4$  dB criterion is still exceeded, a significant site anomaly is present which should be readily apparent and corrective action taken before proceeding to the vertical polarization *NSA* measurement.

### 5.6.1 General *NSA* measurement

For each polarization measurement, the *NSA* procedure requires two different measurements of  $V_R$  which is the voltage received. The first reading of  $V_R$  is with the two coaxial cables disconnected from the two antennas and connected to each other via an adapter. The second reading of  $V_R$  is taken with the coaxial cables reconnected to their respective antennas and the maximum signal measured when the receive antenna is scanned in height. (1 m to 4 m for 3 m and 10 m separation distances and either 1 m to 4 m or 2 m to 6 m for the 30 m separation.) For both of these measurements, the signal source voltage,  $V_i$ , is kept constant. The first reading of  $V_R$  is called  $V_{\text{DIRECT}}$  and the second is  $V_{\text{SITE}}$ . These are used in the following Equation (1) for the measured *NSA*,  $A_N$ ; all terms are in dB.

$$A_N = V_{\text{DIRECT}} - V_{\text{SITE}} - AF_T - AF_R - \Delta AF_{\text{TOT}} \quad (1)$$

where

- $AF_T$  is the transmit antenna factor;
- $AF_R$  is the receive antenna factor;
- $\Delta AF_{\text{TOT}}$  is the mutual impedance correction factor.

Note that the first two terms represent the actual measurement of site attenuation, i.e.,  $V_{\text{DIRECT}} - V_{\text{SITE}}$  is equal to the classical view of site attenuation, which is constituted by the insertion loss of the propagation path with the inclusion of the properties of the two antennas used. Theoretical values for  $\Delta AF_{\text{TOT}}$  are given in Table E.4.  $AF_T$  and  $AF_R$  shall be measured.

Note that:  $V_{\text{DIRECT}} = V_i - C_T - C_R$

where

$C_T$  and  $C_R$  are the cable losses which do not need to be measured separately. The mutual impedance correction factor in Table E.4 applies only to the recommended site geometry of 3 m separation, horizontal polarization and the use of half-wavelength tuned dipoles.

To accomplish these *NSA* measurements, two techniques can be used, depending on the instrumentation available and whether a broadband or tuned dipole is used. Both methods give essentially equal results if used correctly as outlined in Annex E. Briefly, each method is described as follows:

a) Discrete frequency method

For this method, specific frequencies given in Tables E.1, E.2 or E.3 are measured in turn. At each frequency, the receive antenna is scanned over the height range given in the appropriate table to maximize the received signal. These measured parameter values are inserted in Equation (1) to obtain the measured *NSA*. Annex E contains a suggested procedure approach to record the data, calculate the measured *NSA*, and then compare it with the theoretical *NSA*.

b) Swept frequency method

For this method, measurements using broadband antennas may be made using automatic measuring equipment having a peak hold (maximum hold), storage capability, and a tracking generator. In this method, both antenna height and frequency are scanned or swept over the required ranges. The frequency sweep speed shall be much greater than the antenna height scan rate. Otherwise the procedure is the same as in a). A detailed procedure is given in Annex E.

### 5.6.2 Antenna factor determination

Accurate antenna factors are necessary in measuring *NSA*. In general, antenna factors provided with the antenna are inadequate unless they are specifically or individually measured. Linearly polarized antennas are required. A useful antenna calibration method is contained in Annex E. Manufacturer's antenna factors may account for losses due to the balun among other features. If a separate balun or any integrally associated cables are used, their effects must be accounted for. The formula to use for tuned half-wave dipoles is also contained in Annex E.

### 5.6.3 Site attenuation deviations

If measurements of *NSA* deviate by more than  $\pm 4$  dB, several items should be re-checked first:

- a) measurement procedure;
- b) accuracy of antenna factors;
- c) drift in signal source or accuracy of receiver or spectrum analyzer input attenuator and reading.

If no errors are found in a), b) and c), then the site is at fault and detailed investigation of possible causes of site variability should be made. Annex F contains the errors that can occur with *NSA* measurements.

Note that since the vertical polarization is generally the more critical measurement, site anomalies should be investigated using this more sensitive measurement rather than the horizontal polarization results. Key items to investigate include:

- a) ground plane size and construction inadequacy;
- b) objects at the perimeter of the site that may be causing undesired scattering;
- c) all-weather cover;

- d) ground plane discontinuity at the turntable circumference when the turntable surface is conductive and at the same elevation as the ground plane;
- e) thick dielectric ground plane covers;
- f) openings in ground plane for stairways.

## 5.7 Test site suitability with ground-plane

There are many different test sites and facilities that have been constructed to make radiated emission measurements. Most are protected from the weather and the adverse effects of the radio frequency ambient. These include all weather-covered open area test sites and absorber-lined shielded rooms.

Whenever construction material encloses a test site, there is the possibility that the results of a single normalized site attenuation (*NSA*) measurement, as specified in 5.6, are not adequate to show such alternative site suitability.

To assess alternative test site suitability, the following procedure is recommended. It is based on making multiple *NSA* measurements throughout a volume occupied by the EUT. These *NSA* measurements shall all come within the error budget of  $\pm 4$  dB to be judged suitable as an equivalent to an open area test site.

The discussion in this section concerns alternative test sites which have a conducting ground plane.

### 5.7.1 Normalized site attenuation for alternative test sites

For an alternative test site a single *NSA* measurement is insufficient to pick up possible reflections from the construction and/or RF-absorbing material comprising the walls and ceiling of the facility. For these sites a "test volume" is defined as that volume traced out by the largest equipment or system to be tested as it is rotated about its centre location through  $360^\circ$ , such as by a turntable. In evaluating horizontal and vertical polarization, such as illustrated in Figures 6a and 6b, it may require a maximum of 20 separate site attenuation measurements, i.e. five positions in the horizontal plane (centre, left, right, front, and rear, measured with respect to the centre and a line drawn from the centre to the position of the measuring antenna), for two polarizations (horizontal and vertical), and for two heights (1 m and 2 m horizontal, 1 m and 1,5 m vertical).

These measurements are carried out with a broadband antenna and distances are measured with respect to the centre of the antenna. The transmit and receive antennas shall be aligned with the antenna elements parallel to each other and orthogonal to the measurement axis.

For vertical polarization, the off-centre positions of the transmit antenna are at the periphery of the test volume. Furthermore, the lower tip of the antenna shall be greater than 25 cm from the floor, which may require the centre of the antenna to be slightly higher than 1 m for the lowest height measurement.

For horizontal polarization measurements in the left and right positions if the distance between the construction and/or absorbing material on the side walls and EUT periphery is less than 1 m, the centre of the antenna is moved towards to central position so that the extreme tip of the antenna is either at the periphery or distant from the periphery by not more than 10 % of the test volume diameter. The front and rear positions are at the periphery of the test volume.

The number of required measurements can be reduced under the following circumstances.

- a) The vertical and horizontal polarization measurements in the rear position may be omitted if the closest point of the construction and/or absorbing material is at a distance greater than 1 m from the rear boundary of the test volume.

NOTE Radiated emission sources located near dielectric interfaces have been shown to have variations in current distribution that can affect the radiated properties of the source at that location. When EUT can be located near these interfaces, additional site attenuation measurements are required.

- b) The total number of horizontal polarization measurements along the test volume diameter joining the left and right positions may be reduced to the minimum number necessary for the antenna footprints to cover 90 % of the diameter.
- c) The vertical polarization measurements at the 1,5 m height may be omitted if the top of the EUT, including any table mounting, is less than 1,5 m in height.
- d) If the test volume is no larger than 1 m in depth, by 1,5 m in width, by 1,5 m in height, including table if used, horizontal polarization measurements need only be made at the centre, front and rear positions but at the height of both 1 m and 2 m. If item a) above applies, the rear position may be omitted. This will require a minimum of eight measurements: four positions vertical polarization (left, centre, right, and front) for one height, and four positions horizontal polarization (centre and front) for two heights; see Figures 6c and 6d.

NSA measurements shall be performed with the transmit and receive antenna separation held constant according to Tables 1 and 2. Note that these tables have been modified to accommodate these NSA measurements by adding values for an additional transmit height and to limit the 30 m scan height to between 1 m and 4 m. The receive antenna must be moved to maintain the appropriate separation along a line towards the turntable centre (see Figures 6a, 6b, 6c and 6d). The alternative test site is considered suitable for performing radiated emission testing if all NSA measurements prescribed above meet the requirements of 5.7.2 and the ground plane requirements of 5.7.3 below.

NOTE Studies are underway to determine if any further tests are required to show alternate test site suitability.

### 5.7.2 Site attenuation

A measurement site shall be considered acceptable for radiated electromagnetic field measurements if the measured horizontal and vertical NSA measurements are within  $\pm 4$  dB of the theoretical normalized site attenuation for an ideal site.

### 5.7.3 Conducting ground plane

A conducting ground plane is required at a radiated emission test site. The conducting ground plane shall extend at least 1 m beyond the periphery of the EUT and the largest measurement antenna, and cover the entire area between the EUT and the antenna. It shall be of metal with no holes or gaps having longitudinal dimensions larger than one-tenth of a wavelength at the highest frequency of measurement. A larger size conducting ground plane may be required if the NSA measurements do not meet the  $\pm 4$  dB criterion.

NOTE Ongoing studies may indicate the need for specifying minimum conductive ground plane size.

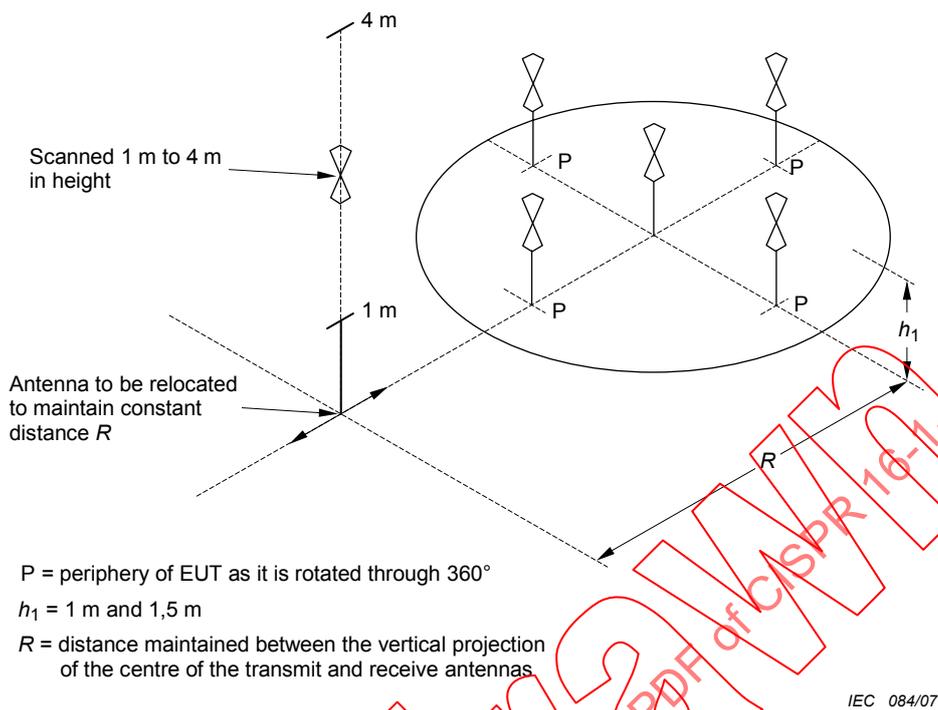


Figure 6a – Typical antenna positions for alternative test site – Vertical polarization NSA measurements

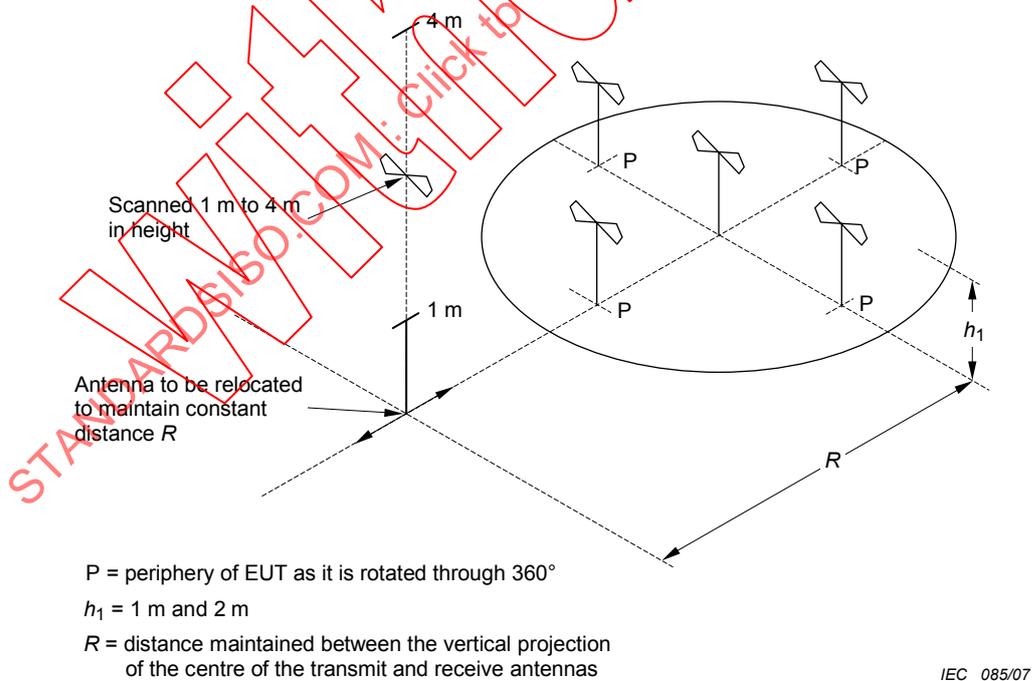


Figure 6b – Typical antenna positions for alternative test site – Horizontal polarization NSA measurements



**Table 1 – Normalized site attenuation  
(recommended geometries for tuned half-wave dipoles with horizontal polarization)**

Polarization	Horizontal	Horizontal	Horizontal
$R$	3 m	10 m	30 m
$h_1$	2 m	2 m	2 m
$h_2$	1 m to 4 m	1 m to 4 m	1 m to 4 m
$f_m$ MHz	$A_N$ dB		
30	11,0	24,1	41,7
35	8,8	21,6	39,1
40	7,0	19,4	36,8
45	5,5	17,5	34,7
50	4,2	15,9	32,9
60	2,2	13,1	29,8
70	0,6	10,9	27,2
80	-0,7	9,2	24,9
90	-1,8	7,8	23,0
100	-2,8	6,7	21,2
120	-4,4	5,0	18,2
140	-5,8	3,5	15,8
160	-6,7	2,3	13,8
180	-7,2	1,2	12,0
200	-8,4	0,3	10,6
250	-10,6	-1,7	7,8
300	-12,3	-3,3	6,1
400	-14,9	-5,8	3,5
500	-16,7	-7,6	1,6
600	-18,3	-9,3	0
700	-19,7	-10,6	-1,4
800	-20,8	-11,8	-2,5
900	-21,8	-12,9	-3,5
1 000	-22,7	-13,8	-4,5

**Table 2 – Normalized site attenuation\*  
(recommended geometries for broadband antennas)**

Polarization	Horizontal	Horizontal	Horizontal	Vertical	Vertical	Vertical	Vertical
$R$	3 m	10 m	30 m	3 m	3 m	10 m	30 m
$h_1$	1 m	1 m	1 m	1 m	1,5 m	1 m	1 m
$h_2$	1 m to 4 m	1 m to 4 m	1 m to 4 m	1 m to 4 m	1 m to 4 m	1 m to 4 m	1 m to 4 m
$f_m$ MHz	$A_N$ dB						
30	15,8	29,8	47,8	8,2	9,3	16,7	26,0
35	13,4	27,1	45,1	6,9	8,0	15,4	24,7
40	11,3	24,9	42,8	5,8	7,0	14,2	23,5
45	9,4	22,9	40,8	4,9	6,1	13,2	22,5
50	7,8	21,1	38,9	4,0	5,4	12,3	21,6
60	5,0	18,0	35,8	2,6	4,1	10,7	20
70	2,8	15,5	33,1	1,5	3,2	9,4	18,7
80	0,9	13,3	30,8	0,6	2,6	8,3	17,5
90	-0,7	11,4	28,8	-0,1	2,1	7,3	16,5
100	-2,0	9,7	27	-0,7	1,9	6,4	15,6
120	-4,2	7,0	23,9	-1,5	1,3	4,9	14,0
140	-6,0	4,8	21,2	-1,8	-1,5	3,7	12,7
160	-7,4	3,1	19	-1,7	-3,7	2,6	11,5
180	-8,6	1,7	17	-1,3	-5,3	1,8	10,5
200	-9,6	0,6	15,3	-3,6	-6,7	1,0	9,6
250	-11,7	-1,6	11,6	-7,7	-9,1	-0,5	7,7
300	-12,8	-3,3	8,8	-10,5	-10,9	-1,5	6,2
400	-14,8	-5,9	4,6	-14,0	-12,6	-4,1	3,9
500	-17,3	-7,9	1,8	-16,4	-15,1	-6,7	2,1
600	-19,1	-9,5	0	-16,3	-16,9	-8,7	0,8
700	-20,6	-10,8	-1,3	-18,4	-18,4	-10,2	-0,3
800	-21,3	-12,0	-2,5	-20,0	-19,3	-11,5	-1,1
900	-22,5	-12,8	-3,5	-21,3	-20,4	-12,6	-1,7
1 000	-23,5	-13,8	-4,4	-22,4	-21,4	-13,6	-3,5

\* This data applies to antennas that have at least 25 cm of ground plane clearance when the centre of the antennas is 1 m above the ground plane in vertical polarization.

### 5.8 Test site suitability without ground-plane

The procedure for test sites without ground-plane in the frequency range 30 MHz to 1 000 MHz is as follows.

### 5.8.1 Measurement considerations for free space test sites, as realized by fully absorber-lined shielded enclosures

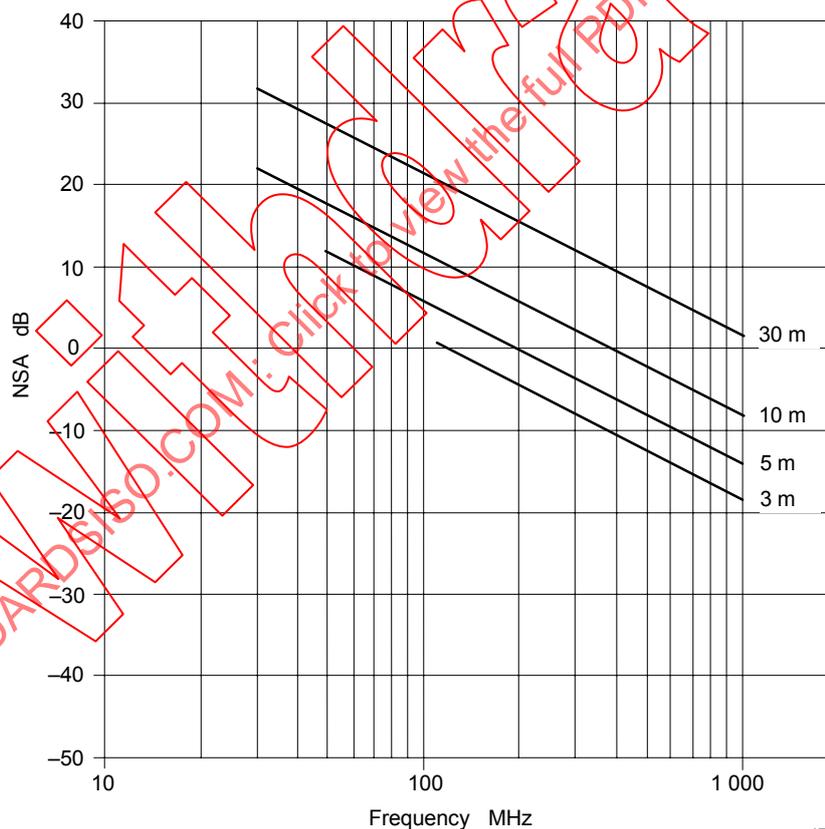
A fully absorber lined shielded enclosure, also known as a fully anechoic chamber (FAC), or a fully anechoic room (FAR), may be used for radiated emission measurements. When the FAR method is used, appropriate radiated emission limits shall be defined in relevant standards (generic, product or product family standards). Compliance with the radio services protection requirements (limits) shall be established for FARs in a similar way as for tests on an OATS.

A FAR is intended to simulate a free space environment such that only the direct ray from the transmitting antenna or EUT reaches the receiving antenna. All indirect and reflected waves shall be minimized with the use of appropriate absorbing material on all walls, the ceiling and the floor of the FAR.

### 5.8.2 Site performance

Site performance may be validated by two methods which are described below – the site reference method and the NSA method.

#### 5.8.2.1 Theoretical normalized site attenuation



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Figure 7 – Graph of theoretical free-space NSA as a function of the frequency for different measurement distances (see Equation 4)

NOTE Frequencies below 110 MHz for 3 m measurement and below 60 MHz for 5 m measurement distances include near field effects. These must be calculated for each individual test site.

The following describes the *NSA* theory for infinitely small antennas.

Site attenuation (*SA*) is the transmission loss measured between the connectors of two antennas on a particular site. For a free space environment, *SA* (in dB) can be approximated by Equation (2)<sup>1)</sup>

$$SA = 20\log_{10} \left[ \left( \frac{5Z_0}{2\pi} \right) \left( \frac{d}{\sqrt{1 - \frac{1}{(\beta d)^2} + \frac{1}{(\beta d)^4}}} \right) \right] - 20\log_{10} f_m + AF_R + AF_T \quad (2)$$

where

$AF_R, AF_T$  are the antenna factors of the receive and transmit antennas in dB/m;

$d$  is the distance between the phase centres of both antennas in metres;

$Z_0$  is the reference impedance (i.e. 50  $\Omega$ );

$\beta$  is defined as  $2\pi/\lambda$ ; and

$f_m$  is the frequency in MHz.

The theoretical normalized site attenuation (*NSA*) in dB is defined as site attenuation with respective antenna factors subtracted, thus:

$$NSA_{\text{calc}} = 20\log_{10} \left[ \left( \frac{5Z_0}{2\pi} \right) \left( \frac{d}{\sqrt{1 - \frac{1}{(\beta d)^2} + \frac{1}{(\beta d)^4}}} \right) \right] - 20\log_{10} f_m \quad (3)$$

Below 60 MHz at a 5 m distance or 110 MHz at a 3 m distance, it is necessary to apply near field correction factors for each of the required test positions of Table 3 for comparison with the theoretical *NSA* of Figure 7 and Equation (2). Near field correction factors are specific to the antennas, test distance, and test volume used, and therefore must be obtained by using a numerical modelling code such as NEC. Alternatively, the site reference method of 5.8.2.2.1 provides cancellation of near field terms if the same antennas and frequencies are used for both the site reference measurement and FAR validation.

For measurement distances of 10 m and 30 m, the near-field terms in Equation (3) may be omitted, and the equation simplifies as follows:

$$NSA_{\text{calc}} = 20\log_{10} \left[ \frac{5Z_0 d}{2\pi} \right] - 20\log_{10} f_m \quad (4)$$

1) Reference: GARBE, H. New EMC Test Facilities for Radiation Measurements. *Review of Radio Science* 1999-2002. John Wiley & Sons, New York, 2002.

If simplified Equation (4) is used instead of Equation (2), the error introduced is less than 0,1dB at frequencies above 60 MHz for 5 m distance and above 110 MHz for 3 m distance. The error will be >0,1 dB below these frequencies due to near-field effects. For a 3 m distance, the maximum error is 1 dB at 30 MHz. To reduce this error Equation (2) should be used.

### 5.8.2.2 Site validation procedure

The NSA shall satisfy the requirement of 5.8.3 over a cylindrical test volume generated by the rotation of the EUT on the turntable. In this context “the EUT” includes all components of a multi-unit EUT and the interconnecting cables. Table 3 defines the maximum height and diameter ( $h_{\max} = d_{\max}$ ) of the test volume as a function of test distance. This ratio between diameter and test distance ensures an acceptable uncertainty in EUT emissions testing.

**Table 3 – Maximum dimensions of test volume versus test distance**

Maximum diameter $d_{\max}$ and height $h_{\max}$ of the test volume m	Test distance $D_{\text{nominal}}$ m
1,5	3,0
2,5	5,0
5,0	10,0

A single position SA (site attenuation) measurement may not be sufficient to pick up possible reflections from the room construction and/or absorbing material lining the walls, floor, ceiling and turntable of the FAR.

The fully anechoic room SA measurements and validation shall therefore be performed at 15 measurement positions for both horizontal and vertical antenna polarizations of the transmit antenna in the test volume (see Figure 8):

- at three heights of the test volume: bottom, middle and top;
- at five positions in all three horizontal planes: centre, left, right, front and rear positions in each horizontal plane. The rear position may be omitted if the distance between rear position and absorbers is more than 0,5 m. During EUT testing, the rear position on the turntable is also turned to the front, and the contribution of the back reflection will then not affect the maximum signal.

For SA measurements two broadband antennas shall be used: one transmit antenna with its reference point at the measurement positions of the test volume and one receive antenna outside this test volume at a prescribed orientation and position. The transmit antenna shall have an approximately omni-directional  $H$ -plane pattern. (The maximum dimension shall not exceed 40 cm for a 3 m test distance; at larger distances, the size of the antenna can be scaled accordingly).

Typical receive antennas are hybrid (biconical/LPD combination) antennas for 30 MHz to 1 000 MHz, or separate [biconical antennas (for 30 MHz to 200 MHz) and LPD antennas (for 200 MHz to 1 000 MHz)].

NOTE Use of a hybrid (biconical/LPD combination) antenna is not recommended for either emission testing or chamber validation at 3 m distance, due to the large physical size of such antennas.

The same antennas, cables, ferrites, attenuators, amplifier, signal generator and receiver used to measure the SA of the FAR, shall be used to measure the reference SA on the quasi-free space test site (5.8.2.2.2). The receive antenna used during the room validation shall be of the same type as used during radiated emission testing of the EUT.

For test volume validation both in horizontal and vertical polarization, and for all transmitting antenna positions in the test volume, the position in height of the receiving antenna in the FAR shall be set and remain at the fixed middle level of the test volume, as shown in Figures 8 and 9. Tilting the antennas shall be necessary to align the bore sight axis of both antennas in one measurement axis. The distance between the antenna reference point (defined in antenna calibration) and the front position of the test volume is  $d_{nominal}$ . When the transmit antenna is moved to other positions in the test volume, the receive antenna shall be translated along the measurement axis to maintain  $d_{nominal}$ . The measurement axis is the line between the transmit and receive antenna, along which  $d_{nominal}$  is defined. For all positions and polarizations, the receiving and the transmitting antenna must face one another with the elements of both antennas parallel (tilting, see Figure 9). Any antenna masts and the supporting floors shall be in place during the validation procedure.

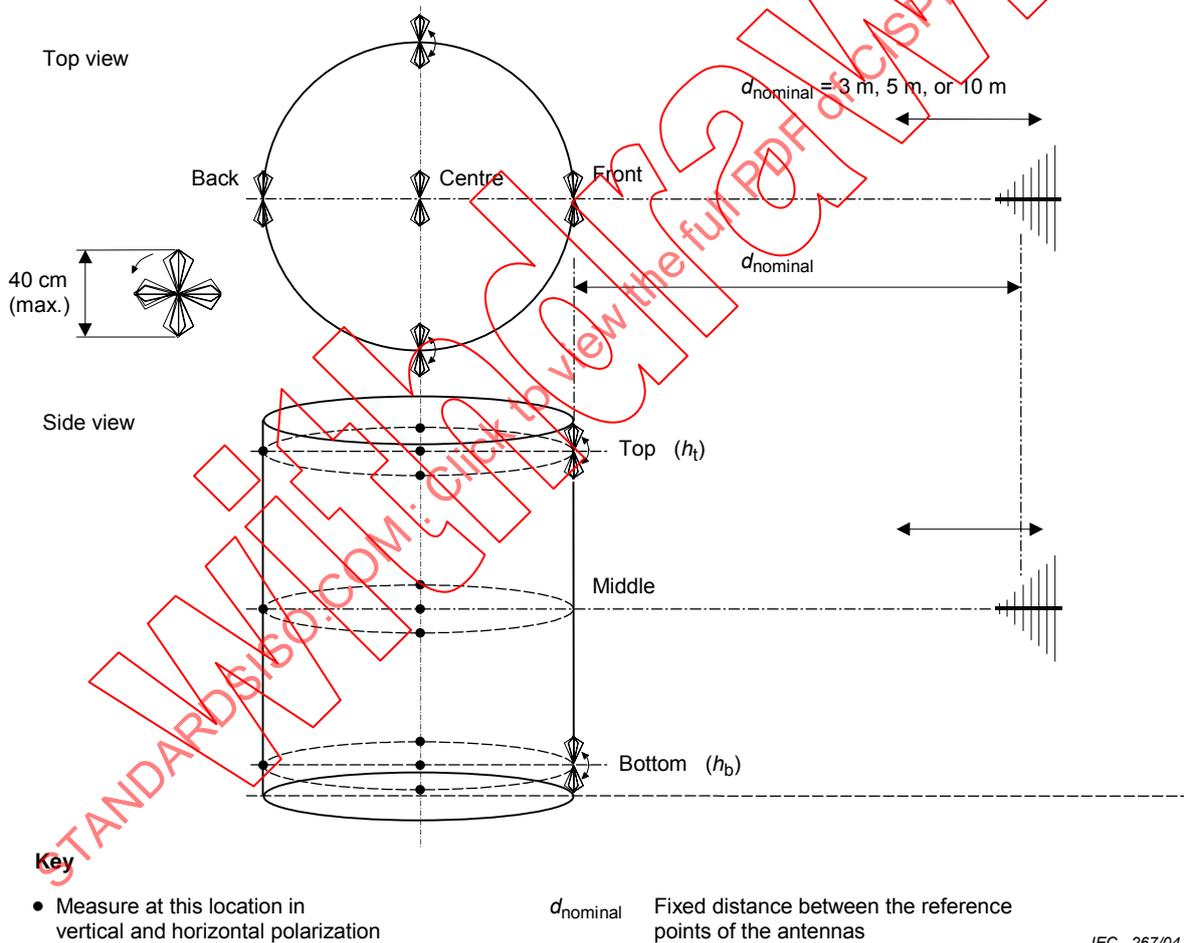


Figure 8 – Measurement positions for the site validation procedure

For all positions of the transmitting antenna in the test volume, in both horizontal and vertical polarizations, the transmitting and receiving antennas shall be aligned on the measurement axis.

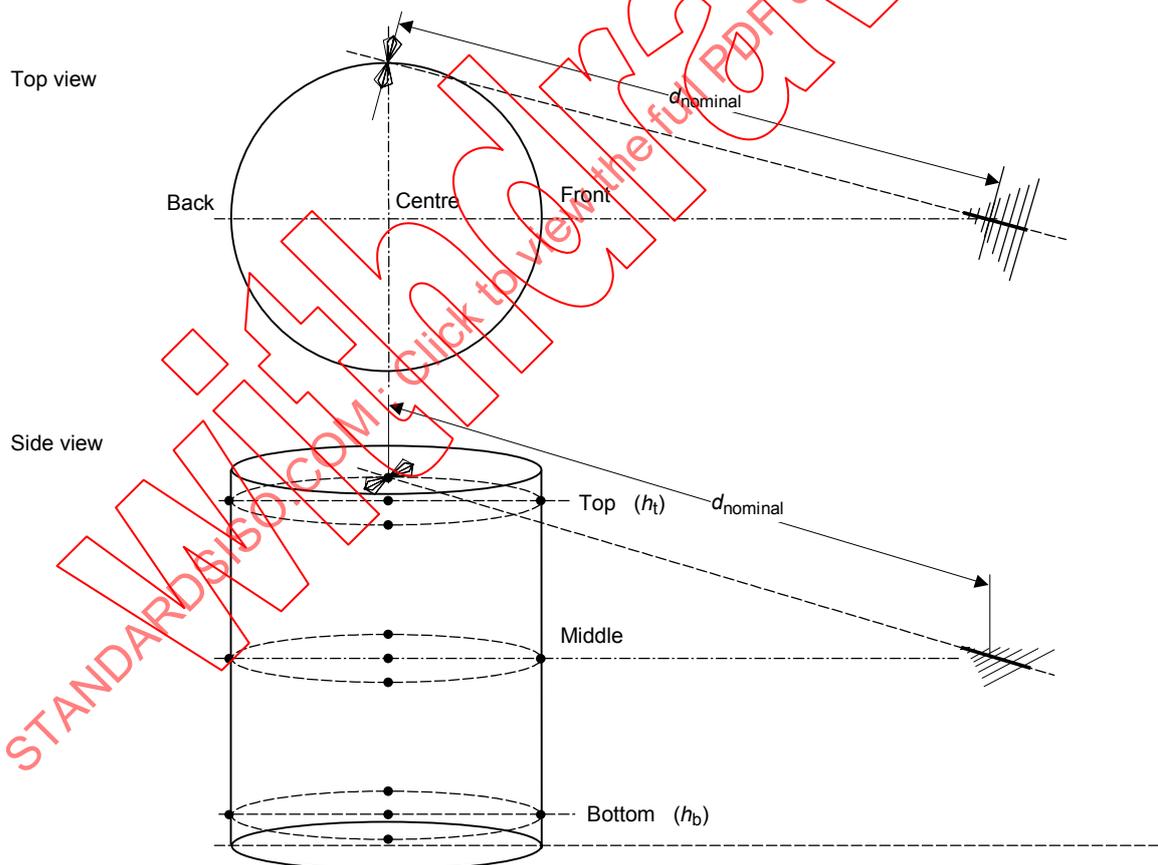
Tilting the antennas is necessary to meet this requirement at certain positions (see Figure 9).

$d_{\text{nominal}}$  { is the test distance associated with the limit;  
is the fixed antenna distance in the validation procedure;  
is the antenna separation in the antenna calibration procedure.

The transmit antenna height position in the test volume shall be determined as follows:

- "Middle" where possible along a virtual axis positioned at mid-height and mid-width of the FAR;
- "top ( $h_t$ )" and "bottom ( $h_b$ )" by half of  $h_{\text{max}}$  (see Table 3) minus half of the transmit antenna dimension (e.g. 20 cm for small biconical antenna).

These adjusted positions shall be used for both vertical and horizontal polarizations. The distance between the top and bottom planes and the ceiling and floor absorbers respectively is given by the absorber performance as determined by the volumetric NSA test, but at least 0,5 m, to avoid EUT to absorber coupling.



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NOTE Antenna polarization horizontal, position top right.

**Figure 9 – Example of one measurement position and antenna tilt for the site validation procedure**

The maximum step size for the discrete-frequency measurement shall be as listed in Table 4:

**Table 4 – Frequency ranges and step sizes**

Frequency range MHz	Maximum frequency step MHz
30 – 100	1
100 – 500	5
500 – 1 000	10

Two methods are permissible for site validation:

- a) the site reference method, which is required for test distances less than 5 m;
- b) the *NSA* method, which is preferred for test distances greater than or equal to 5 m.

The *SA* measurement methods are intended to provide 0 dB deviation when performed on an ideal site. Any methods may be implemented to decrease measurement uncertainty, as long as these do not contradict the defined set-up and procedures or hide any site deficiencies, e.g. smoothed resonances.

Site validation measurement uncertainty can be decreased by the following measures.

- For a vertically-polarized antenna, shielded cables are to be extended by at least 2 m behind each antenna before dropping the cable to the ground. If possible, cables shall extend straight back to the bulkhead connectors in the wall of the room. Another possibility is the use of clip-on ferrites on the cables. Another alternative for reducing the influence of cables is through the use of optical links.
- Attenuators at the antenna connectors (e.g. 6 dB or 10 dB) will reduce the influence of any large impedance mismatch at the antennas.
- Antennas with good balance of the balun shall be used (the receiver reading changes less than  $\pm 0,5$  dB when the antenna is rotated through  $180^\circ$  with respect to its bore sight axis. Antenna balance verification methods are described in 4.4.2).
- Separate biconical and LPD antennas for chamber evaluation may be used (antenna type is changed at 200 MHz); if these will be used for EUT testing. A hybrid (biconical/LPD combination ) antenna is a combination of these two types and may be used as well if the mechanical dimensions are sufficiently small for the measurement distance.

The FAR site validation procedure shall be performed at regular intervals, to detect long-term changes in room characteristics, and when changes that might influence the electromagnetic wave transmission characteristics in the fully anechoic room occur.

### 5.8.2.2.1 The site reference method

SA measurements with the antenna pair (transmit and receive antenna) on a quasi-free space test site are required as a reference. The procedure for determining this reference site attenuation ( $SA_{ref}$ ) is described in 5.8.2.2.2. This method accounts for mutual coupling of the antennas and near field effects, which can have a significant influence at 3 m test distances. The reference site attenuation  $SA_{ref}(d)$  is performed at the nominal distance,  $d_{nominal}$ , between the transmit and receive antennas.

The site validation procedure for each test volume position is performed in three steps.

- 1)  $M0$  is the reference level measured by the receiver in  $\text{dB}\mu\text{V}$  with the cables connected together, normally done once before a series of volumetric tests.
- 2)  $M1$  is the level measured by the receiver in  $\text{dB}\mu\text{V}$  with antennas installed.

The site attenuation of the validated site  $SA_{val}$  can be calculated by

$$SA_{val} = M0 - M1 \quad \text{in dB} \quad (5)$$

- 3) The deviation of the measured site attenuation ( $\Delta SA$ ) from reference site attenuation  $SA_{ref}(d)$  is calculated using Equation (6).

$$\Delta SA = SA_{ref}(d) - SA_{val} \quad \text{in dB} \quad (6)$$

### 5.8.2.2.2 Determining the site reference

For accurate site validations at distances less than 5 m, it is recommended that dedicated pairs of antennas be used to determine the site reference (transmit and receive antenna). A quasi-free space test site is required. It consists of 2 non-metallic antenna masts (wood or plastic with  $\epsilon_r \leq 2,5$ , low loss, diameter as small as possible retaining mechanical strength), which allow the placement of antennas at a certain height above the ground level (Figure 10). One possible method of realization of the  $\pm 1$  dB performance of the reference site is to choose the height ( $h$ ) of the antennas as follows

$$h \geq d \times 8/3 \quad (7)$$

where  $d$  is the antenna separation.

A height of  $h = d \times 8/3$  is recommended to suppress the influence of the ground, or substantial absorbers which work down to 30 MHz, need to be placed on the ground.

NOTE At 3 m separation at 30 MHz there is a significant near field term ( $1/d^2$ ) that alone contributes an error of 0,8 dB for a height of 5/3. This was verified by the national laboratories of both the UK and Austria. For a site reference with an uncertainty of less than  $\pm 0,5$  dB, a height of 8/3 is recommended if no absorber is placed at the ground.

The distance shall be equal to the actual distance  $d_{nominal}$  between the antennas used in the FAR. The antennas are polarized vertically (horizontal polarization shall not be used because of stronger interference with the ground-reflected signal). It also provides a good approximation of free-space. The clearance from buildings, trees, etc. shall be greater than  $d \times 8/3$  because there may be an influence for vertically polarized antennas.

Care has to be taken that the antenna feed cables do not affect the test result. This is best avoided by a cable arrangement as shown in Figure 10, or using RF-optical links.

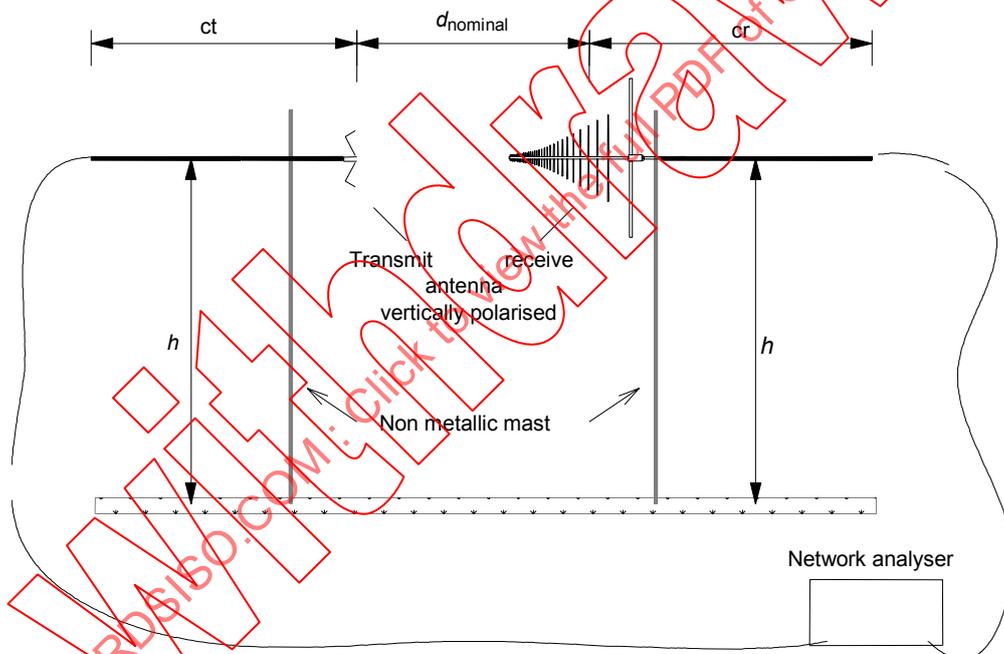
The quality of the reference set-up directly influences the FAR evaluation result.

The site reference ( $SA_{\text{ref}}$ ) is determined in 3 steps, as follows.

- 1)  $M0_{\text{RS}}$  is the reference level measured by the receiver in  $\text{dB}\mu\text{V}$  with the cables connected together.
- 2)  $M1_{\text{RS}}(d)$  is the level measured by the receiver in  $\text{dB}\mu\text{V}$  with the antennas installed at the required distance  $d_{\text{nominal}}$ .
- 3) The  $SA_{\text{ref}}(d)$  is calculated according to Equation (8)

$$SA_{\text{ref}}(d) = M0_{\text{RS}} - M1_{\text{RS}}(d) \text{ in dB} \quad (8)$$

For 3 m site validation a height of at least 4 m above ground shall be used, which is a typical capability of remotely controllable antenna masts that are used for emission measurements. In this case, electromagnetic absorbers shall be placed on the ground between the antennas, with the absorber patch extending for a minimum area beyond the antennas in all directions and it must be proven that quasi-free space condition, as defined in 5.8.1, is fulfilled. For site validation with  $d > 3$  m, the equation  $h > d \times 8/3$  is used, or an alternative set-up that has been demonstrated to fulfil the  $\pm 1$  dB reference site attenuation.



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#### Key

$d_{\text{nominal}}$  validation distance

$h$  height of the antennas above a ground plane or above ground level

ct, cr coaxial feed cables for transmit and receive antenna oriented horizontally behind the antenna for a distance as close to 2 m as physically possible. In a FAR, route the cables horizontally as far as possible, preferably straight through a hole in the chamber wall, or use optical fibre connected to an RF-optical link on the output of the antenna.

NOTE Site reference is obtained separately for all geometries of Figure 10.

**Figure 10 – Typical free-space site reference measurement set-up**

### 5.8.2.2.3 The NSA method

The free space antenna factors of the transmit and receive antenna (defined by antenna calibration clauses of the CISPR 16 series) are required for this procedure. The site validation for each measurement position is performed in 4 steps as follows.

- 1)  $M_0$  is the reference level measured by the receiver with the cables connected together.
- 2)  $M_1$  is the level measured by the receiver with the antennas installed.
- 3) The measured NSA ( $NSA_m$ ) is calculated in dB according to Equation (9)

$$NSA_m = M_0 - M_1 - AF_T - AF_R \text{ in dB} \quad (9)$$

where  $AF_T$  and  $AF_R$  are free space antenna factors in dB/m.

- 4) The deviation  $\Delta NSA$  is calculated in dB according to Equation (10)

$$\Delta NSA = NSA_m - NSA_{calc} \quad (10)$$

where  $NSA_{calc}$  is calculated using Equation (4), and  $\Delta NSA$  is compared with the applicable NSA criterion, e.g.  $\pm 4$  dB, as specified in 5.8.3.

NOTE The distance  $d$  between the reference points of the transmit and receive antennas (defined by antenna calibration) must be used as  $d_{nominal}$ . The effective distance between the antennas varies with frequency due to their phase centre positions. The transmission loss shall be compensated by the ratio of the effective distance to  $d_{nominal}$ .

### 5.8.3 Site validation criteria

A measurement site shall comply with the following requirements:

- deviations of the SA or the NSA (Equation 6 or Equation 10) shall be less than  $\pm 4$  dB for both horizontal and vertical polarization and for each measurement position and measurement frequency,
- the uncertainty budget of the site evaluation according to CISPR 16-4-2 recommendations must be reported and shall have the same components as required for field strength measurements on alternative test sites with ground plane.

## 5.9 Evaluation of set-up table and antenna tower

### 5.9.1 Introduction

A set-up table as specified in Clause D.5 typically positions the EUT for field strength measurements. The shape, construction and material permittivity of the set-up table can influence the field strength measurement results (see Bibliography). The following subclause (5.9.2) describes a procedure to determine the influence of the set-up table for the 30 MHz to 1 000 MHz frequency range and to estimate its related uncertainty contribution to field strength measurements.

NOTE Only horizontal polarisation of a transmit antenna above the setup table is used in the evaluation. This polarisation accounts for the worst-case effects from the table.

The antenna tower does not require additional evaluation because any perturbation effects will be included in the NSA measurement.

### 5.9.2 Evaluation procedure for set-up table influences (table-top equipment)

The type, shape, and component materials of a set-up table may affect the field strength measurement results. An evaluation procedure shall be performed to determine these effects, and to estimate the standard uncertainties caused by the table. To evaluate set-up table influences, two transmission measurements are performed with a specific transmit antenna in a specific arrangement with and without the set-up table. This difference between the measurement results with and without the set-up table gives an estimate of the influence caused by the set-up table. The measurement procedure is as follows:

The set-up table shall be placed in the typical position on the test site with the largest dimension (i.e. the diagonal for a set-up table with a rectangular top, or the radius for a table with circular a top) facing the receive antenna direction (see Figure 11). For the frequency range up to 1 000 MHz, a small biconical antenna with an overall length of less than 0,40 m is placed above the set-up table in horizontal polarization. The distance between the surface of the set-up table and the balun centre is 0,1 m (see Figure 12). The small biconical antenna is positioned with the reference point (balun) midway between the centre and the edge of the setup table top in the direction of the receive antenna. A signal generator feeds the transmit antenna above the set-up table. The frequency steps shall be less than or equal to 0,5 % of the highest frequency used. The receive antenna voltage shall be at least 20 dB above the noise level of the measurement equipment. The feed cable is routed horizontally to the rear for approximately 2 m at the same height as the antenna. Ferrite tubes should be placed on the receive antenna feed cable at suitable intervals to prevent the feed cable from influencing measurements.

Two transmission measurements shall be performed to investigate the maximum voltage  $V_r$  at the receive antenna, with transmit antenna position unchanged for each test – one with and one without the set-up table. In the frequency range below 1 GHz, measurements shall be performed at least in the frequency range of 200 MHz<sup>2)</sup> to 1 GHz. The receive antenna is height scanned between 1 m and 4 m for an OATS or SAC, while in a FAR, the receive antenna is at a fixed height.

The difference  $\Delta(f)$  between the two measurement results is then calculated using the Equation (11), with measured voltages are in dB( $\mu$ V),

$$\Delta(f) = |V_{r/with} - V_{r/without}| \quad (11)$$

where

$V_{r/without}$  is voltage measured at a specific frequency without the set-up table;

$V_{r/with}$  is voltage measured at a specific frequency with the set-up table.

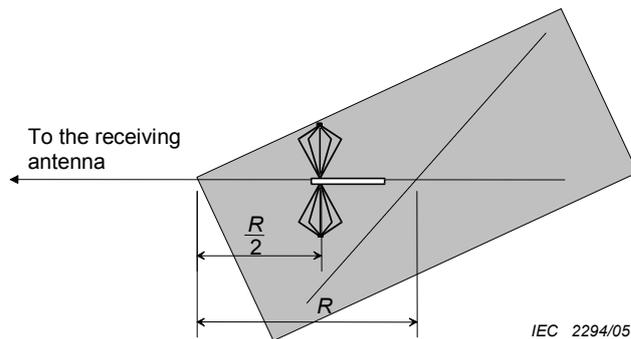
The magnitude of the maximum difference  $\Delta_{max}$  in the frequency range of 200 MHz to 1 000 MHz is used as the estimated maximum deviation, with  $\Delta_{max}$  is in dB,

$$\Delta_{max} = \max |V_{r/with} - V_{r/without}|_{200 \text{ MHz} - 1 \text{ 000 MHz}} \quad (12)$$

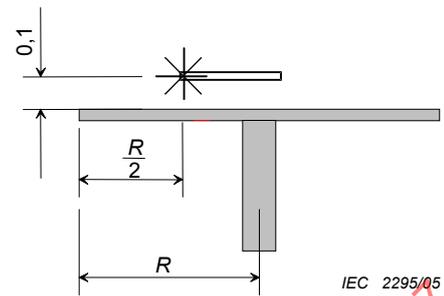
The standard uncertainty  $u_{table}$  caused by the set-up table is estimated by assuming a rectangular distribution for the measured maximum difference  $\Delta_{max}$ . So  $u_{table}$  (in dB) can be calculated using Equation (13).

$$u_{table} = \frac{1}{\sqrt{3}} \cdot \Delta_{max} \quad (13)$$

2) Below about 200 MHz, the influence of the set-up table is negligible when applying this verification procedure.



**Figure 11 – Position of the antenna relative to the edge above a rectangle set-up table (top view)**



**Figure 12 – Antenna position above the set-up table (side view)**

NOTE Test laboratories typically may apply different types of tables of which the construction and the type of materials may differ. It is sufficient to determine the worst-case value of  $\Delta$  (or  $V_{r,with}$ ) in the determination of  $U_{table}$ .

### 5.9.3 Evaluation procedure for set-up table influences (floor standing equipment)

The set-up table for a floor-standing EUT shall be constructed using a non-conducting, low-permittivity material. If the set-up table perimeter is less than or equal to the EUT perimeter at the base (footprint), evaluation of table is not required.

## 6 Reverberating chamber for total radiated power measurement

For some types of equipment operating in the microwave frequency range, because of the existence of complex three-dimensional radiation patterns which are sensitive to equipment operating conditions and its surroundings, the measurement of total radiated power is considered to be a significant parameter related to disturbance control. It can be measured by placing the equipment in a suitable chamber with metal walls. To avoid effects of standing waves that would otherwise produce non-uniform distribution of energy density with position in the chamber, rotating stirrers are installed. With proper size, shape and position, the energy density at any position in the chamber varies randomly with a constant statistical distribution law in phase, amplitude and polarization.

### 6.1 Chamber

#### 6.1.1 Size and shape

The linear dimensions of the chamber shall be large relative to the wavelength of the lowest frequency of interest. It shall also be large enough to accommodate the equipment under test, the stirrers and the measuring antennas. Microwave equipment varies in size from the small table top oven having a volume of about 0,2 m<sup>3</sup> to large units 1,7 m high with a 760 mm base. The chamber may be of any shape provided its three dimensions are of the same order. The three dimensions should preferably be different. For a lowest frequency of 1 GHz, the chamber shall have a volume at least 8 m<sup>3</sup>. The actual dimensions will depend on the physical characteristics of the chamber. See 6.1.4 for method of test of the suitability of the chamber.

The walls and the stirrers shall be metallic. Joints between the metallic members shall be mechanically sound and of low electrical resistance along the whole length, and there shall be no surface corrosion. No absorbing material, such as wood, shall be placed inside the chamber.

### 6.1.2 Door, openings in walls, and mounting brackets

The enclosure door shall be large enough to allow the passage of operators and equipment. It shall open outward, and fit tightly to minimize energy losses. For convenience in mounting, transmitting and receiving antennas inside the chamber, mounting brackets may be fixed to the walls.

### 6.1.3 Stirrers

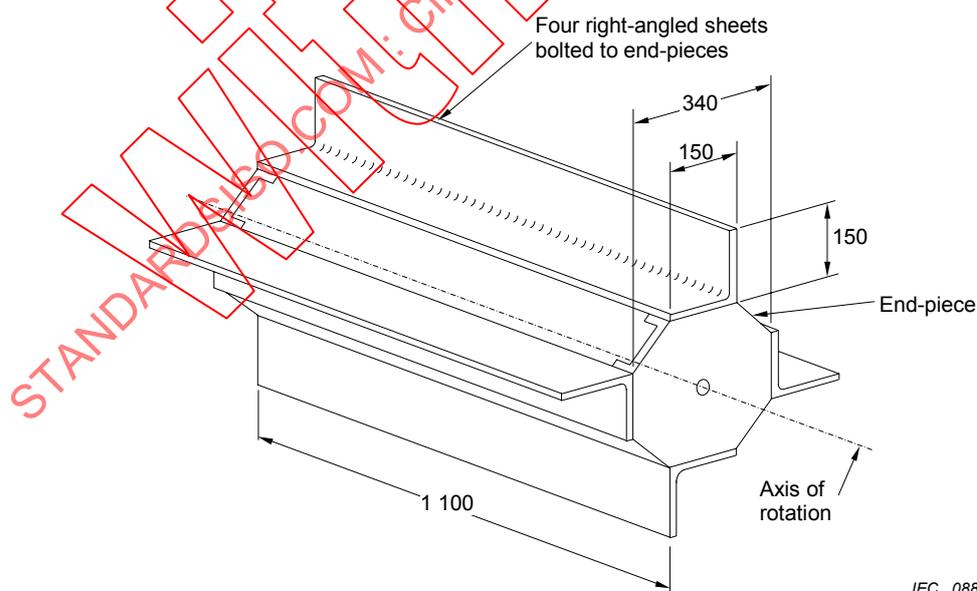
The following describes two examples of stirrers. Other shapes are permissible provided stirring efficiency meets the criteria in 6.1.4.

#### 6.1.3.1 Rotating vanes

If rotating vanes are used, two vanes are placed on adjacent walls of the chamber spaced at least  $1/4$  of the maximum wavelength used from the walls and of sufficient thickness to be rigid. They shall be of the maximum length allowed by the wall sizes and their width shall be about  $1/5$  of the length.

#### 6.1.3.2 Rotating paddles

If rotating paddles are used, two or three paddles are mounted on the walls of the chamber. The paddles shall be mutually at right angles. The paddles may be of the shape shown in Figure 13 and rotate about an axis parallel to their length. The diameter of the swept tubular space shall be at least equal to the maximum wavelength used, and the lengths shall be the maximum allowed by the wall sizes. The structure shall be rigid.



IEC 088/07

Dimensions in millimetres

Figure 13 – Example of a typical paddle stirrer

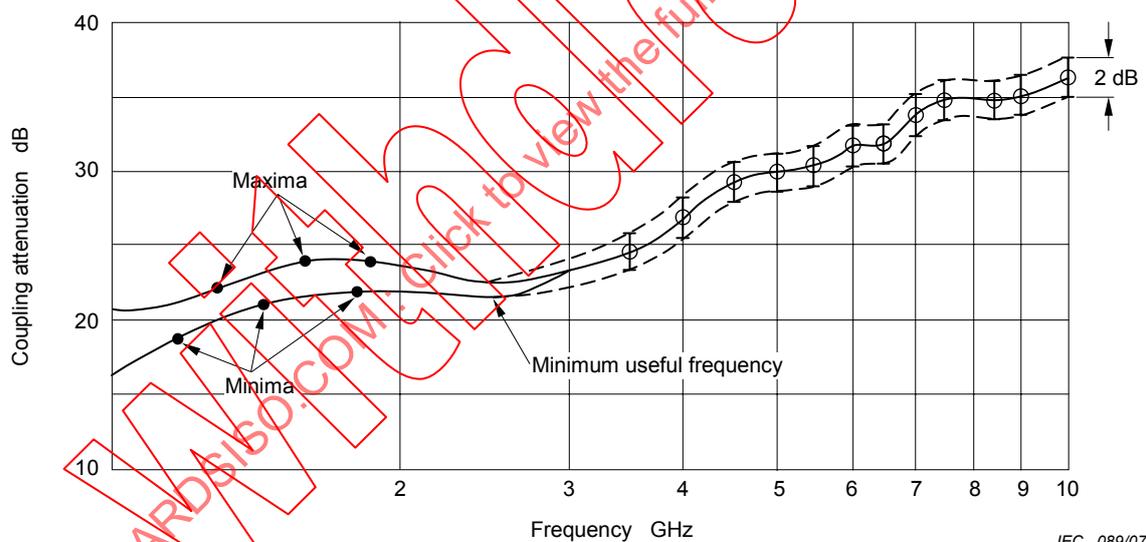
### 6.1.3.3 Rotating speed

The rotation speeds of the stirrers shall be different. The longest time for one rotation of the stirrers shall be less than 1/5 of the integrating time of the measuring instrument. For the measuring equipment described in 6.1.5, a suitable rate is between 50 r/min and 200 r/min. The motors used to rotate the stirrers, together with their reduction gear, should preferably be outside the walls of the chamber.

### 6.1.4 Test for the efficiency of the stirrers

The desired uniform distribution of energy in the chamber is shown by the smoothness of the variation with frequency of coupling attenuation (described in 6.1.5). At low frequencies, due to the longer wavelengths, it is more difficult to achieve this uniformity and there exist pronounced maxima and minima. The greater the efficiency of the stirrers the smaller are these maxima and minima and hence the usable frequency is lower.

The coupling attenuation is measured over the usable frequency range of the chamber. At the lower frequencies where the maxima and minima are observable, values shall be measured at about 100 MHz intervals. The receiving antenna then remains fixed, the transmitting antenna is rotated at 45-degree intervals and the test is repeated for each position and at each frequency. The whole test shall be repeated again with the receiving antenna rotated at 90°. The stirrers are considered satisfactory when: (1) the envelope of the graph of the maxima and the minima does not exceed 2 dB in any position of the transmitting antenna, and, (2) the means of the four graphs are within an envelope of 2 dB or less. Figure 14 shows a typical result.



NOTE All measured points should lie inside the 2 dB envelope marked by the dotted line.

**Figure 14 – Range of coupling attenuation as a function of frequency for a chamber using the stirrer in Figure 13**

### 6.1.5 Coupling attenuation

The coupling attenuation of a chamber is the insertion loss measured between the terminals of the transmitting and the receiving antennas in the chamber. A calibrated signal generator whose power output can be accurately measured is used to feed power to a low-loss transmitting antenna (e.g. a horn antenna) located inside the chamber or on a chamber wall. A receiving antenna may be placed at any point in the chamber provided it is at least 1/4 wavelength from the walls and not pointing toward the transmitting antenna, towards the nearest chamber wall, or aligned with any of the chamber axis.

A low-noise RF amplifier is connected to the receiving antenna via a high-pass filter; its output is connected through a band-pass filter to a diode detector. The band-pass filter shall be tuned to the frequency of interest and be of the specified bandwidth. The output of the detector is connected to a peak reading voltmeter with a specified peak-hold time (the hold time will depend on the equipment being measured). A spectrum analyzer may also be used for this measurement. The power absorbed by the transmitting antenna,  $P$ , is noted. The signal generator is then connected to the input of the low-noise amplifier, and its power output,  $p$ , is adjusted to give the same voltmeter reading. The power absorbed by the low-noise amplifier is noted. The coupling attenuation is  $10 \log (P/p)$  dB.

## 7 TEM cells for immunity to radiated disturbance measurement

(Under consideration)

## 8 Test sites for measurement of radio disturbance field strength for the frequency range 1 GHz to 18 GHz

The test site shall rely on reflection-free conditions. It may be necessary to use absorbing material and/or to raise the height of the EUT to achieve these free-space conditions.

NOTE In the case of floor standing equipment tests, reflection-free conditions may not be achieved close to the ground.

### 8.1 Reference test site

The reference test site shall be a free-space, open area test site (FSOATS) with precautions to ensure that reflections do not influence the measurement.

### 8.2 Validation of the test site

A test site shall be considered acceptable for radiated electromagnetic field measurements in 1 GHz to 18 GHz if it satisfies the criterion provided in 8.2.1; 8.2.2 provides the site validation procedure. For the purposes of testing per CISPR standards, site validation measurements shall be performed from 1 GHz to the maximum frequency in use at the test facility; the maximum frequency shall be at least 2 GHz.

Test sites used for measurements in 1 GHz to 18 GHz shall have a design that minimizes the influence of reflections upon the received signal, for example an anechoic chamber. If the site is not designed to provide fully-anechoic conditions, for example a semi-anechoic chamber, use of absorbing material to cover part of the metal ground plane is required, as described below.

In cases where the test volume extends from the conducting floor of the facility to above the EUT, as may be typical for facilities used primarily for testing floor-standing EUTs, absorber shall be placed in the test volume for the validation as necessary. To accommodate testing of floor-standing equipment which cannot be positioned above the ground plane, illumination of the test volume for a height of up to 30 cm may be obstructed by absorber placed on the ground plane.

During the emission testing of a floor-standing EUT, floor absorber used during the site validation may be removed in the immediate area (footprint) of the EUT, and for up to 10 cm surrounding the EUT footprint.

In facilities where the test volume is above the height of the absorber, as may be typical of facilities used for testing table-top equipment, absorber may be placed under the test volume for both site validation and equipment tests. Photographs showing the site absorber configuration and transmit/receive antenna locations shall be included in the site validation report.

Site validation is performed by measurements of the so-called site voltage standing-wave ratio ( $S_{VSWR}$ ). The site validation method evaluates a given test volume for the specific combination of site, receive antenna, test distance (described in Subclause 7.3.6.1 of CISPR 16-2-3), and absorbing material placed on the ground plane, if needed to meet the criterion of 8.2.1. Influences of the receive antenna mast located as used for the site validation tests, and permanently-fixed objects in the test volume (such as a permanently-installed turntable), are evaluated by and included in this site validation procedure. Removable objects, such as a removable test table, are not required to be in place during the site validation tests if their influence is to be evaluated separately using the additional procedures of 5.8 of this standard.

CISPR 16-2-3 provides a description of the EUT measurement method used for testing in 1 GHz to 18 GHz. The purpose of the  $S_{VSWR}$  procedure is to check for the influence of reflections that may be incident upon an EUT of arbitrary size and shape placed within the test volume as evaluated using this procedure.

The  $S_{VSWR}$  is the ratio of maximum received signal to minimum received signal, caused by interference between direct (intended) and reflected signals, or

$$S_{VSWR} = \frac{E_{\max}}{E_{\min}} = \frac{V_{\max}}{V_{\min}} \quad (14)$$

where  $E_{\max}$  and  $E_{\min}$  are the maximum and minimum received signals, and  $V_{\max}$  and  $V_{\min}$  are the corresponding measured voltages when a receiver or spectrum analyzer is used for reception.

For the procedures that follow, decibels (dB) are typically employed for measurements and calculations. In this case,  $S_{VSWR}$  is given by

$$S_{VSWR,dB} = 20 \log \left( \frac{V_{\max}}{V_{\min}} \right) = 20 \log \left( \frac{E_{\max}}{E_{\min}} \right) = V_{\max,dB} - V_{\min,dB} = E_{\max,dB} - E_{\min,dB} \quad (15)$$

NOTE 1 When decibels are employed,  $S_{VSWR,dB}$  may be taken as the difference of maximum to minimum signal received in units of dBm, dB $\mu$ V, or dB $\mu$ V/m, as appropriate for the instrumentation or signal detector used.

NOTE 2 The value of  $S_{VSWR}$  or  $S_{VSWR,dB}$  is computed separately from the maximum and minimum signal obtained at each frequency and polarization for a set of six measurements as described in 8.2.2.

### 8.2.1 Acceptance criterion for site validation

The  $S_{VSWR}$  is directly related to influences of undesired reflections. The acceptance criterion for 1 GHz to 18 GHz site validations is:

$$S_{VSWR} \leq 2:1, \text{ or } S_{VSWR,dB} \leq 6,0 \text{ dB},$$

for  $S_{VSWR}$  measured in accordance with the procedures of 8.2.2.

### 8.2.2 Site validation procedures

This subclause describes the required procedures for evaluating  $S_{VSWR}$ .

#### 8.2.2.1 Antenna requirements

To provide illumination of all reflecting surfaces during this test, and to simulate the possible low-directivity antenna gains exhibited by many actual EUTs, this subclause specifies characteristics for equipment used for  $S_{VSWR}$  testing. Manufacturer-supplied data may be used to evaluate whether the test-equipment requirements are met.

##### 8.2.2.1.1 Test equipment for the standard $S_{VSWR}$ procedure (8.2.2.3)

The receive antenna must be linearly polarized, and shall be the same type as used for EUT emissions measurements. For the transmit antenna, the 0°-reference angle for the pattern specifications is the angle where the antenna faces the receive antenna (aperture planes parallel); this is also deemed the "bore-sight" direction,  $B$ .

The antenna used as a transmit source shall be linearly polarized and shall have a dipole-like radiation pattern with the following detailed characteristics. Radiation pattern data shall be available with a frequency step size less than or equal to 1 GHz.<sup>3)</sup>

##### 8.2.2.1.1.1 Transmit antenna E-plane radiation pattern

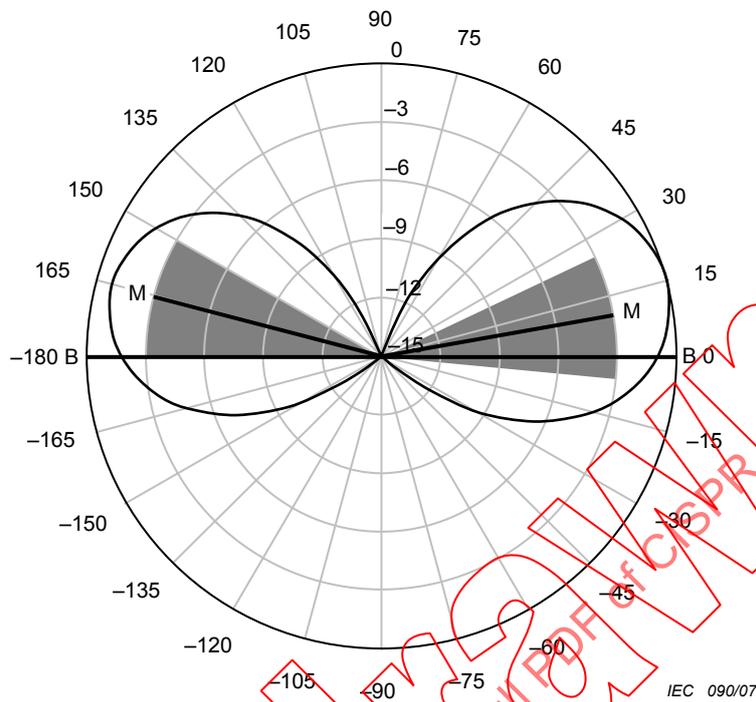
An E-plane radiation pattern for an antenna with simple linear polarization can be measured at one of many possible cut planes (constant azimuth angle) around the radiation sphere. The cut plane for pattern measurements shall be selected by the antenna manufacturer and described in the antenna characterization report. One convenient choice typically is the plane containing the connector and the cable routing.

- a) Choose a main lobe direction, designated as  $M$ , for the right and the left side of each pattern.  $M$  shall be between  $0^\circ \pm 15^\circ$  and  $180^\circ \pm 15^\circ$ , respectively.
- b) Draw the so-called forbidden area symmetrical to the main lobe directions on both sides of the pattern<sup>4)</sup> where amplitude is  $\leq -3$  dB for  $\pm 15^\circ$ .
- c) The E-plane pattern shall not enter the forbidden area.

<sup>3)</sup> It is assumed that the antenna also fulfils the requirements at other frequencies used for the  $S_{VSWR}$  test.

<sup>4)</sup> This limit ensures a smooth pattern in the bore-sight region, and an acceptable omni-directional behaviour.

Figure 15 shows an example radiation pattern that meets the preceding E-plane requirements.



NOTE The example plot is for an antenna that meets the E-plane requirements of 8.2.2.1.1.1. The main lobe directions, *M*, for the right and the left side of each pattern are between  $0^\circ \pm 15^\circ$  and  $180^\circ \pm 15^\circ$  respectively. The shaded areas represent the "forbidden area" where amplitude would be  $\leq -3$  dB for  $\pm 15^\circ$  of each main lobe. The antenna pattern does not enter the forbidden area.

**Figure 15 – Transmit antenna E-Plane radiation pattern example (for informative purposes only)**

**8.2.2.1.1.2 Transmit antenna H-plane radiation pattern**

There is only one possible plane in which to measure the H-plane pattern of a dipole antenna, which is the plane orthogonal to the dipole axis intersecting the centre of the dipole. This plane may include a balun, an input connector, and the input cable, depending whether a metal or optical fiber is used. The manufacturer of the antenna shall describe the set-up used to measure radiation patterns, including the feed cabling and connector locations, in the test report of the antenna.

- a) Average the radiation pattern data (in dB) over the range of  $\pm 135^\circ$  ( $0^\circ$  is the bore-sight angle). The maximum step size for this pattern data is  $5^\circ$  in the frequency range of 1 GHz to 6 GHz, and  $1^\circ$  from 6 GHz to 18 GHz.
- b) The pattern must not exceed the following deviations from the  $\pm 135^\circ$ -averaged value:

Angle range	1 GHz to 6 GHz	6 GHz to 18 GHz
$-60^\circ$ to $60^\circ$	$\pm 2$ dB	$\pm 3$ dB
$-60^\circ$ to $-135^\circ$ , $60^\circ$ to $135^\circ$	$\pm 3$ dB	$\pm 4$ dB
$-135^\circ$ to $-180^\circ$ , $135^\circ$ to $180^\circ$	$< +3$ dB	$< +4$ dB

NOTE Although a lower bound on the H-plane pattern is not specified outside of  $\pm 135^\circ$ , it is desirable for the H-plane pattern not to show a null at  $\pm 180^\circ$ , but to be omni-directional as best as possible. Guidance provided by the antenna manufacturer on the routing of the feed cabling and antenna mast should be followed, if available, to minimize the possible influence on H-plane pattern outside of  $\pm 135^\circ$ .

Figure 16 shows an example pattern that meets the preceding H-plane requirements.

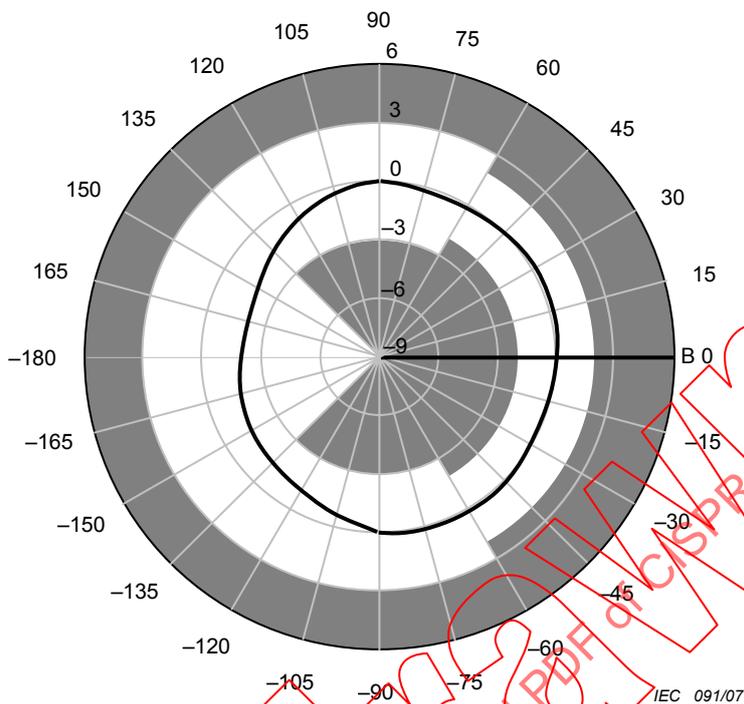


Figure 16a – 1 GHz to 6 GHz

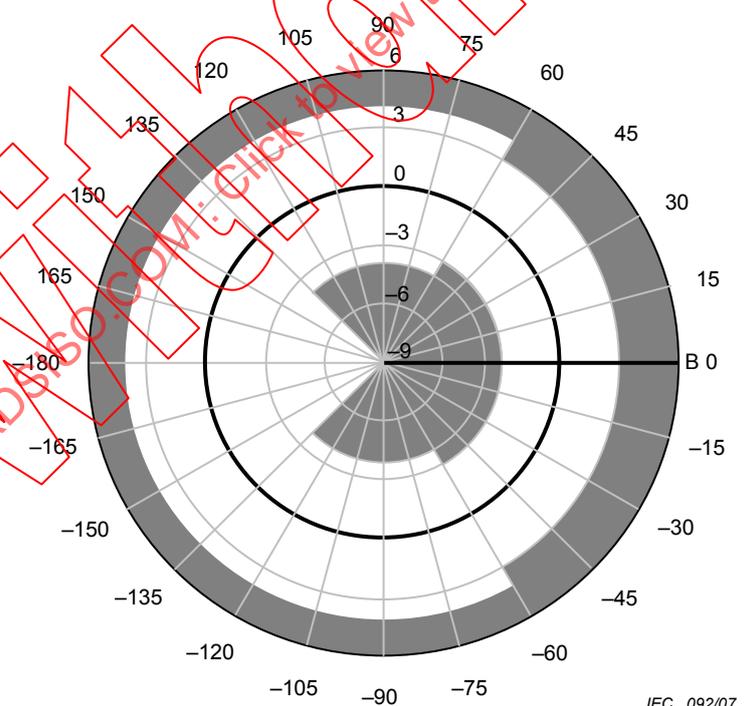


Figure 16b – 6 GHz to 18 GHz

NOTE The example plot is for an antenna that meets the H-plane requirements. The shaded areas represent the maximum permissible deviations stated in 8.2.2.2.1.1.2. This example antenna meets the requirements because the pattern does not enter the shaded regions.

Figure 16 – Transmit antenna H-plane radiation pattern (for informative purposes only)

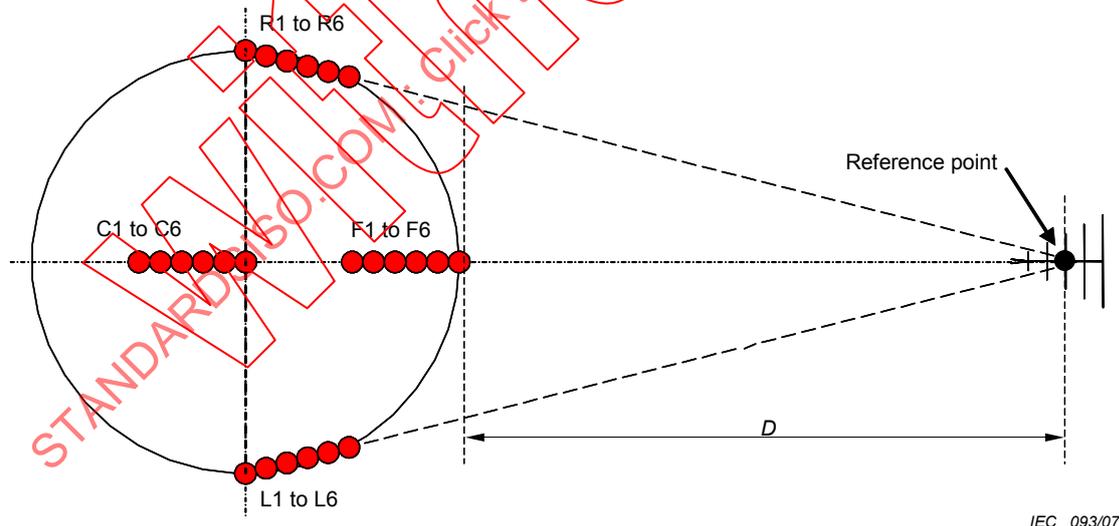
### 8.2.2.1.2 Test equipment for the reciprocal $S_{VSWR}$ procedure (8.2.2.4)

The antenna used to transmit from the test volume shall be the same type as used later for emissions measurements. The isotropic field probe used shall be omni-directional with an isotropicity of 3 dB or better.

### 8.2.2.2 Required positions for site validation testing

The site validation test shall be performed for a volume in the shape of a cylinder. The bottom of the cylinder is established by the surface that is used to support the EUT. The top of the cylinder is chosen as the maximum height that an EUT and its vertical overhead cabling would occupy. The diameter of the cylinder is the largest diameter required to accommodate an EUT including cables. For cables that leave the test volume, a 30 cm section of these cables shall be assumed to establish the dimensions of the volume. To accommodate floor-standing equipment that cannot be raised above the supporting surface, test-volume illumination for a height of up to 30 cm from the bottom of the test volume is allowed to be obstructed by absorber placed on the ground plane. According to the procedure of 8.2.2.3, the  $S_{VSWR}$  is evaluated by placing the receive antenna at the position for which the volume shall be validated, and varying the transmit source location across the defined positions. Alternatively, using the reciprocal  $S_{VSWR}$  procedure of 8.2.2.4, the positions described in this subclause are used for the placement of the field probe in the test volume.

The required locations to perform the  $S_{VSWR}$  measurements are dependent upon the dimensions of the test volume. Details of the conditional test position requirements are given in 8.2.2.5. The  $S_{VSWR}$  is evaluated for each required location and polarization by a sequence of six measurements along a line to the reference point of the receive antenna. All of the possible required locations are illustrated in Figure 17 and Figure 18, including the conditional locations described in 8.2.2.5. The sequence of six measurements along the line to the receive antenna is indicated by dots in these figures.



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Figure 17 –  $S_{VSWR}$  measurement positions in a horizontal plane – see 8.2.2.2.1 for description

### 8.2.2.2.1 Descriptions of $S_{VSWR}$ measurement positions in a horizontal plane (Figure 10)

- a) Front positions 1-6 (F1 to F6): The front positions are on a line from the center of the test volume to the receive antenna reference point. To locate these positions, first locate F6 at the front extent of the test volume, on the measurement axis spaced away at the test distance,  $D$ , from the reference point of the receive antenna.

F5 to F1 are measured relative to F6 as follows, moving away from the receive antenna:

- 1)  $F5 = F6 + 2$  cm away from the receive antenna
- 2)  $F4 = F6 + 10$  cm away from the receive antenna
- 3)  $F3 = F6 + 18$  cm away from the receive antenna
- 4)  $F2 = F6 + 30$  cm away from the receive antenna
- 5)  $F1 = F6 + 40$  cm away from the receive antenna

- b) Right positions 1-6 (R1 to R6): These positions are located relative to position R6. R6 is found by determining the right extent of the test volume (position R1), then moving on a line toward the receive antenna reference point 40 cm (see Figure 17).

Positions R5 to R1 are measured relative to R6 as follows, moving away from the receive antenna:

- 1)  $R5 = R6 + 2$  cm away from the receive antenna
- 2)  $R4 = R6 + 10$  cm away from the receive antenna
- 3)  $R3 = R6 + 18$  cm away from the receive antenna
- 4)  $R2 = R6 + 30$  cm away from the receive antenna
- 5)  $R1 = R6 + 40$  cm away from the receive antenna

- c) Left positions 1-6 (L1 to L6): These positions are located relative to position L6. L6 is found by determining the left extent of the test volume (position L1), then moving on a line toward the receive antenna reference point 40 cm (see Figure 17).

Positions L5 to L1 are measured relative to L6 as follows, moving away from the receive antenna:

- 1)  $L5 = L6 + 2$  cm away from the receive antenna
- 2)  $L4 = L6 + 10$  cm away from the receive antenna
- 3)  $L3 = L6 + 18$  cm away from the receive antenna
- 4)  $L2 = L6 + 30$  cm away from the receive antenna
- 5)  $L1 = L6 + 40$  cm away from the receive antenna

- d) Center positions 1-6 (C1 to C6): These positions are located relative to position C6. Position C6 is at the center of the test volume. Positions C1 to C6 are required to be tested when the test volume diameter is greater than 1,5 m (see 8.2.2.5).

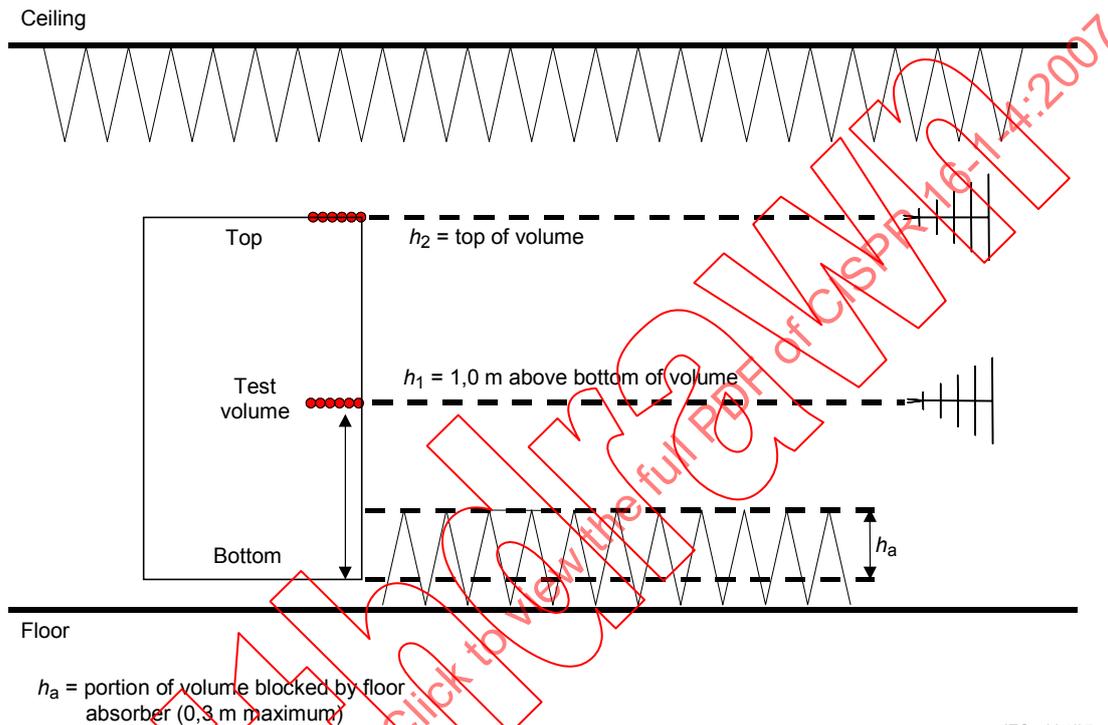
C5 to C1 are measured relative to C6, moving away from the receive antenna as follows:

- 1)  $C5 = C6 + 2$  cm away from the receive antenna
- 2)  $C4 = C6 + 10$  cm away from the receive antenna
- 3)  $C3 = C6 + 18$  cm away from the receive antenna

- 4)  $C_2 = C_6 + 30$  cm away from the receive antenna
- 5)  $C_1 = C_6 + 40$  cm away from the receive antenna

### 8.2.2.2.2 Descriptions of $S_{VSWR}$ additional measurement positions (Figure 18)

In addition to the locations indicated in Figure 17, an additional  $S_{VSWR}$  test plane at the top of the test volume may be required depending upon the height of the test volume. Figure 18 illustrates the additional height requirement for  $S_{VSWR}$  measurements. The test at the second height is to be performed at the front position only.



#### Key

- $h_a$  the portion of the test volume that is obstructed by absorber placed on the floor (30 cm maximum)
- $h_1$  height located at the middle of the test volume, or 1,0 m above the bottom of the test volume, whichever is lower.
- $h_2$  height located at the top of the test volume and required to be tested when  $h_2$  is separated by at least 0,5 m from  $h_1$  (see 8.2.2.5 for details).

Figure 18 –  $S_{VSWR}$  positions (height requirements)

Table 5 provides a summary of the test positions. In Table 5, the positions are grouped according to height ( $h_1$ ,  $h_2$ ) and location (front, left, right, center). For each location, a reference position is designated for use in the calculations required by Equation (16).

**Table 5 –  $S_{VSWR}$  test positions**

Position name	Location	Height	Polarization	Reference position for $D_{ref}$ [see Equation (13)]	Location relative to reference position
<b>Front positions (Front, <math>h_1</math>) at first height</b>					
F1h1H	Front	$h_1$	Horizontal	F6h1	+40 cm away from the receive antenna
F1h1V	Front	$h_1$	Vertical	F6h1	+40 cm away from the receive antenna
F2h1H	Front	$h_1$	Horizontal	F6h1	+30 cm away from the receive antenna
F2h1V	Front	$h_1$	Vertical	F6h1	+30 cm away from the receive antenna
F3h1H	Front	$h_1$	Horizontal	F6h1	+18 cm away from the receive antenna
F3h1V	Front	$h_1$	Vertical	F6h1	+18 cm away from the receive antenna
F4h1H	Front	$h_1$	Horizontal	F6h1	+10 cm away from the receive antenna
F4h1V	Front	$h_1$	Vertical	F6h1	+10 cm away from the receive antenna
F5h1H	Front	$h_1$	Horizontal	F6h1	+2 cm away from the receive antenna
F5h1V	Front	$h_1$	Vertical	F6h1	+2 cm away from the receive antenna
F6h1H	Front	$h_1$	Horizontal	F6h1	= Reference position (Front, $h_1$ )
F6h1V	Front	$h_1$	Vertical	F6h1	= Reference position (Front, $h_1$ )
<b>Center positions (Center, <math>h_1</math>) at first height (if required - see 8.2.2.5)</b>					
C1h1H	Center	$h_1$	Horizontal	C6h1	+40 cm away from the receive antenna
C1h1V	Center	$h_1$	Vertical	C6h1	+40 cm away from the receive antenna
C2h1H	Center	$h_1$	Horizontal	C6h1	+30 cm away from the receive antenna
C2h1V	Center	$h_1$	Vertical	C6h1	+30 cm away from the receive antenna
C3h1H	Center	$h_1$	Horizontal	C6h1	+18 cm away from the receive antenna
C3h1V	Center	$h_1$	Vertical	C6h1	+18 cm away from the receive antenna
C4h1H	Center	$h_1$	Horizontal	C6h1	+10 cm away from the receive antenna
C4h1V	Center	$h_1$	Vertical	C6h1	+10 cm away from the receive antenna
C5h1H	Center	$h_1$	Horizontal	C6h1	+2 cm away from the receive antenna
C5h1V	Center	$h_1$	Vertical	C6h1	+2 cm away from the receive antenna
C6h1H	Center	$h_1$	Horizontal	C6h1	= Reference position (Center, $h_1$ )
C6h1V	Center	$h_1$	Vertical	C6h1	= Reference position (Center, $h_1$ )

Table 5 (continued)

Position name	Location	Height	Polarization	Reference position for $D_{ref}$ [see Equation (13)]	Location relative to reference position
<b>Right positions at first height</b>					
R1h1H	Right	$h_1$	Horizontal	R6h1	+40 cm away from the receive antenna at the right extent of the volume.
R1h1V	Right	$h_1$	Vertical	R6h1	+40 cm away from the receive antenna at the right extent of the volume.
R2h1H	Right	$h_1$	Horizontal	R6h1	+30 cm away from the receive antenna
R2h1V	Right	$h_1$	Vertical	R6h1	+30 cm away from the receive antenna
R3h1H	Right	$h_1$	Horizontal	R6h1	+18 cm away from the receive antenna
R3h1V	Right	$h_1$	Vertical	R6h1	+18 cm away from the receive antenna
R4h1H	Right	$h_1$	Horizontal	R6h1	+10 cm away from the receive antenna
R4h1V	Right	$h_1$	Vertical	R6h1	+10 cm away from the receive antenna
R5h1H	Right	$h_1$	Horizontal	R6h1	+2 cm away from the receive antenna
R5h1V	Right	$h_1$	Vertical	R6h1	+2 cm away from the receive antenna
R6h1H	Right	$h_1$	Horizontal	R6h1	= Reference position (Right, $h_1$ )
R6h1V	Right	$h_1$	Vertical	R6h1	= Reference position (Right, $h_1$ )
<b>Left positions at first height</b>					
L1h1H	Left	$h_1$	Horizontal	L6h1	+40 cm away from the receive antenna at the left extent of the volume.
L1h1V	Left	$h_1$	Vertical	L6h1	+40 cm away from the receive antenna at the left extent of the volume.
L2h1H	Left	$h_1$	Horizontal	L6h1	+30 cm away from the receive antenna
L2h1V	Left	$h_1$	Vertical	L6h1	+30 cm away from the receive antenna
L3h1H	Left	$h_1$	Horizontal	L6h1	+18 cm away from the receive antenna
L3h1V	Left	$h_1$	Vertical	L6h1	+18 cm away from the receive antenna
L4h1H	Left	$h_1$	Horizontal	L6h1	+10 cm away from the receive antenna
L4h1V	Left	$h_1$	Vertical	L6h1	+10 cm away from the receive antenna
L5h1H	Left	$h_1$	Horizontal	L6h1	+2 cm away from the receive antenna
L5h1V	Left	$h_1$	Vertical	L6h1	+2 cm away from the receive antenna
L6h1H	Left	$h_1$	Horizontal	L6h1	= Reference position (Left, $h_1$ )
L6h1V	Left	$h_1$	Vertical	L6h1	= Reference position (Left, $h_1$ )

Table 5 (continued)

Position name	Location	Height	Polarization	Reference position for $D_{ref}$ [see Equation (13)]	Location relative to reference position
<b>Front positions at second height (if required – see 8.2.2.5)</b>					
F1h2H	Front	$h_2$	Horizontal	F6h2	+40 cm away from the receive antenna
F1h2V	Front	$h_2$	Vertical	F6h2	+40 cm away from the receive antenna
F2h2H	Front	$h_2$	Horizontal	F6h2	+30 cm away from the receive antenna
F2h2V	Front	$h_2$	Vertical	F6h2	+30 cm away from the receive antenna
F3h2H	Front	$h_2$	Horizontal	F6h2	+18 cm away from the receive antenna
F3h2V	Front	$h_2$	Vertical	F6h2	+18 cm away from the receive antenna
F4h2H	Front	$h_2$	Horizontal	F6h2	+10 cm away from the receive antenna
F4h2V	Front	$h_2$	Vertical	F6h2	+10 cm away from the receive antenna
F5h2H	Front	$h_2$	Horizontal	F6h2	+2 cm away from the receive antenna
F5h2V	Front	$h_2$	Vertical	F6h2	+2 cm away from the receive antenna
F6h2H	Front	$h_2$	Horizontal	F6h2	= Reference position (Front, $h_2$ )
F6h2V	Front	$h_2$	Vertical	F6h2	= Reference position (Front, $h_2$ )
NOTE These $S_{VSWR}$ measurements may be performed in any sequence.					

### 8.2.2.3 $S_{VSWR}$ site validation – standard test procedure

In the following procedure, the positions are designated as  $P_{mnopq}$ , where the subscripts correspond to the position names as listed in the first column of Table 5. The measured signal,  $M$ , is the received E-field or voltage measurement at each position, and is similarly denoted by subscripts as  $M_{mnopq}$ . For example,  $P_{F1h1H}$  is the position F1, at height 1, horizontal polarization, and its measured signal (in dB) is referred to as  $M_{F1h1H}$ .

- Locate the transmit source with its reference point at front position 6, height 1, in horizontal polarization ( $P_{F6h1H}$ ). Locate the receive antenna, also in horizontal polarization, at the test distance  $D$ , measured from the source to the reference point of the receive antenna. Note that the receive antenna height shall be located at the same height as the transmit source for all measurements.
- Verify that the received signal displayed will be at least 20 dB above the ambient and above the measuring receiver or spectrum analyzer displayed noise across the entire frequency range to be measured. If not, it may be necessary to use different equipment (antennas, cables, signal generator, preamplifier) and/or use partial frequency ranges as appropriate to maintain a level of 20 dB above the displayed noise floor.
- Record the measured signal level,  $M_{F6h1H}$  at each frequency. Swept measurement or stepped frequency increments may be used. If stepped increments are used, the frequency increment shall be 50 MHz or less.
- Repeat steps a) and b) with the transmit source at the other five positions shown in Table 6 (8.2.2.6) for the front, height 1, horizontal polarization. In total, there will be six measurements for front, height 1, horizontal polarization ( $M_{F1h1H}$  through  $M_{F6h1H}$ ) varying in separation distance from the receive antenna by the increments shown in Table 5.

- e) Change the polarization of the transmit source and receive antenna to vertical and repeat the above procedure for positions  $P_{F1h1V}$  through  $P_{F1h6V}$  in order to obtain  $M_{F1h1V}$  through  $M_{F6h1V}$ .
- f) For all measurements, normalize the measured E-field or voltage data to the distance of the reference position shown in Table 5 using Equation (16):

$$M'_{mnopq} = M_{mnopq} + 20 \log \left( \frac{D_{mnopq}}{D_{ref}} \right) \text{ (dB)} \quad (16)$$

where  $D_{mnopq}$  is the actual separation distance for the measurement location,  $D_{ref}$  is the separation distance measured to the reference position, and  $M_{mnopq}$  is the measured signal (E-field or receiver voltage) in decibels. Note that each measurement location has a different reference position corresponding to position 6, as indicated in Table 5 for  $P_{mnopq}$ .

- g) Using Equation (14) or Equation (15), calculate the  $S_{VSWR}$  for horizontal polarization. Using Equation (15),  $S_{VSWR,dB}$  can be obtained by subtracting the minimum received signal,  $M_{min,dB}$ , from the maximum received signal,  $M_{max,dB}$ , after distance corrections have been applied [step f)] for the six positions. Repeat the calculation for readings obtained using vertical polarization.
- h) The  $S_{VSWR}$  for each polarization shall fulfil the acceptance criteria of 8.2.1.
- i) Repeat steps a)-h) for the left and right positions of the test volume. Note that when the transmit source antenna is moved to the left or right, its bore-sight direction shall be aimed towards the receive antenna. However, the receive antenna shall remain facing towards the center (not aimed at the side positions), which is the same direction it will be facing later during measurements performed on EUTs.
- j) If required by 8.2.2.5, repeat the above procedure for the measurements at the center position, and for the measurements required at the second height. When measurements are performed at the second height, the receive antenna shall be at the same height as the transmit antenna.

#### 8.2.2.4 $S_{VSWR}$ site validation – reciprocal test procedure using an isotropic field probe

For shielded facilities (i.e., fully-anechoic or semi-anechoic chambers), it is permitted to evaluate  $S_{VSWR}$  using an isotropic field probe placed at the required locations of Table 5 and illuminating the test volume with the same antenna that is used later as the receive antenna for emissions testing. For the purposes of this standard, this method is termed the “reciprocal” method of  $S_{VSWR}$  determination. In this  $S_{VSWR}$  reciprocal procedure, the antenna to later be used as the receive antenna in EUT emissions testing is termed the “transmit” antenna, because it will be used to transmit to a probe located in the test volume. The isotropic field probe is required to fulfil the radiation pattern specifications of 8.2.2.1. The probe shall be capable of being aligned with the polarization of the transmit antenna, i.e., the location and orientation of the sensing elements within the probe must be known.

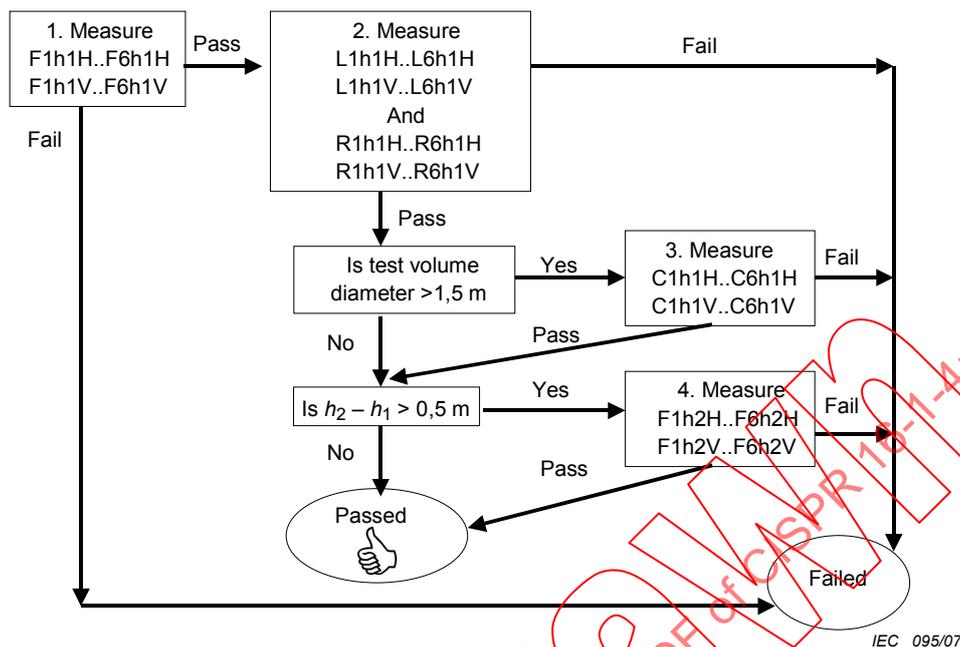
The reciprocal  $S_{VSWR}$  site validation test procedure using an isotropic field probe is as follows.

- a) Place the field probe at the front position 6, height 1, in horizontal polarization ( $P_{F6h1H}$ ). Place the transmit antenna at the test distance  $D$  as measured from the perimeter of the test volume to the reference point of the antenna. The transmit antenna height shall be at the same height as the probe for all positions.

- b) Verify that field strength magnitude is sufficient to allow proper functioning of the probe. For guidance on the equipment and procedures necessary to establish appropriate field strengths, refer to the manufacturers operating specifications for the probe (adequate sensitivity and measurement uncertainty). In addition, the transmit system and probe system should be checked for linearity, and harmonics shall be suppressed to a level of at least 15 dB below the primary signal. Use of a directional coupler is recommended to monitor forward power during the test, because variations in the output power level will produce variations in the test results. It is important to provide stable output signals, because any signal variation due to instability of the signal source (e.g., bad cable connections, variations with warm-up time of the preamplifier, etc.) will result in additional variations of the results (i.e., artificially high  $S_{VSWR}$  results).
- c) Record the measured signal level,  $M_{F6h1H}$ , at each frequency. Swept measurement or stepped frequency increments may be used. If stepped increments are used, the frequency increment shall be 50 MHz or less.
- d) Repeat step c) with the field probe at the other five positions shown in Table 6 (8.2.2.6) for the front, height 1, horizontal polarization. In total, there will be six measurements for front, height 1, horizontal polarization ( $M_{F1h1H}$  through  $M_{F6h1H}$ ) varying in separation distance from the receive antenna by the increments shown in Table 5.
- e) Change the polarization of the field probe and antenna to vertical, and repeat the above procedure for positions  $P_{F1h1V}$  through  $P_{F6h1V}$ , in order to obtain  $M_{F1h1V}$  through  $M_{F6h1V}$ .
- f) For all measurements, normalize the obtained data using Equation (16).
- g) Using Equation (14) or Equation (15), calculate the  $S_{VSWR}$  for horizontal polarization. Using Equation (15),  $S_{VSWR,dB}$  can be obtained by subtracting the minimum received signal,  $M_{min,dB}$ , from the maximum received signal,  $M_{max,dB}$ , after distance corrections have been applied [step f)] for the six positions. Repeat the calculation for the readings obtained using vertical polarization.
- h) The  $S_{VSWR}$  for both polarizations shall fulfil the acceptance criteria of 8.2.1.
- i) Repeat the above procedure for the left and right positions of the test volume. Note that for this reciprocal  $S_{VSWR}$  procedure the probe may be adjusted to maintain a constant direction facing at the reference point of the transmit antenna. However, the transmit antenna shall remain facing toward the center of the volume (not aimed at the side positions) in the same direction it will be facing during later measurements of EUTs.
- j) If required by 8.2.2.5, repeat the above procedure for the measurements at the center position, and for any measurements required at the second height. When measurements are performed at the second height, the probe shall be at the same height as the transmit antenna.

#### 8.2.2.5 Conditional test position requirements

As indicated in Figure 17, Figure 18 and Table 5, additional test positions are required to be tested depending upon the size of the test volume. Figure 19 presents a flow chart specifying when these additional measurements are required.



NOTE The measurements are not required to be performed in the sequence shown, and may proceed in any order such that all the required data is obtained.

Figure 19 – Conditional test position requirements

When additional test positions are required,  $S_{VSWR}$  is to be determined at each test frequency from each group of six measurements independently for horizontal and vertical polarization using the procedures of 8.2.2.3 or 8.2.2.4.

8.2.2.6  $S_{VSWR}$  site validation test report

Table 6 lists a summary of all of the possible required  $S_{VSWR}$  measurements and calculations, including the results from the required positions and the conditional positions of 8.2.2.5.

**Table 6 –  $S_{VSWR}$  reporting requirements**

Location	Height	Polarization	Type	$S_{VSWR}$ dB
Front	$h_1$	Horizontal	Required	$= \text{Max} (S'_{F1h1H} \dots S'_{F6h1H}) - \text{Min} (S'_{F1h1H} \dots S'_{F6h1H})$
Front	$h_1$	Vertical	Required	$= \text{Max} (S'_{F1h1V} \dots S'_{F6h1V}) - \text{Min} (S'_{F1h1V} \dots S'_{F6h1V})$
Right	$h_1$	Horizontal	Required	$= \text{Max} (S'_{R1h1H} \dots S'_{R6h1H}) - \text{Min} (S'_{R1h1H} \dots S'_{R6h1H})$
Right	$h_1$	Vertical	Required	$= \text{Max} (S'_{R1h1V} \dots S'_{R6h1V}) - \text{Min} (S'_{R1h1V} \dots S'_{R6h1V})$
Left	$h_1$	Horizontal	Required	$= \text{Max} (S'_{L1h1H} \dots S'_{L6h1H}) - \text{Min} (S'_{L1h1H} \dots S'_{L6h1H})$
Left	$h_1$	Vertical	Required	$= \text{Max} (S'_{L1h1V} \dots S'_{L6h1V}) - \text{Min} (S'_{L1h1V} \dots S'_{L6h1V})$
Center	$h_1$	Horizontal	Conditional	$= \text{Max} (S'_{C1h1H} \dots S'_{C6h1H}) - \text{Min} (S'_{C1h1H} \dots S'_{C6h1H})$
Center	$h_1$	Vertical	Conditional	$= \text{Max} (S'_{C1h1V} \dots S'_{C6h1V}) - \text{Min} (S'_{C1h1V} \dots S'_{C6h1V})$
Front	$h_2$	Horizontal	Conditional	$= \text{Max} (S'_{F1h2H} \dots S'_{F6h2H}) - \text{Min} (S'_{F1h2H} \dots S'_{F6h2H})$
Front	$h_2$	Vertical	Conditional	$= \text{Max} (S'_{F1h2V} \dots S'_{F6h2V}) - \text{Min} (S'_{F1h2V} \dots S'_{F6h2V})$

The preceding  $S_{VSWR}$  calculations and reporting requirements apply for each test frequency.

### 8.2.2.7 Limitations of the $S_{VSWR}$ site validation method

The measurement points chosen for 8.2.2.2 and contained in the preceding procedures are intended to provide an overall measure of the  $S_{VSWR}$  of the test site across the frequency range of 1 GHz to 18 GHz. Note however that the peak  $S_{VSWR}$  may not always be captured using the procedures of 8.2.2.3 or 8.2.2.4 at any specific frequency  $f$ . Therefore, statements about  $S_{VSWR}$  compliance based on measurements at any single frequency should be avoided. However, the peak found by the above procedures within adjacent octaves ( $0,5f$  to  $2f$ ) is typically representative of the worst case  $S_{VSWR}$  for all frequencies inclusive in the band.

In cases where more accuracy of the  $S_{VSWR}$  result is desired at a single frequency, the above method can be improved by measuring more than six locations along the lines shown Figure 17 and Figure 18. The additional data collection points should be spaced unequally, and chosen based on a distance translation of the source antenna (or field probe in the reciprocal  $S_{VSWR}$  method) using quarter-wavelength steps at the frequency of interest.

### 8.3 Alternative test site

Any measurement site that achieves free-space conditions is a possible alternative test site.

## **Annex A** (normative)

### **Parameters of broadband antennas**

#### **A.1 Introduction**

As new and improved antennas are used in making both radiated emission and immunity measurements over wide frequency ranges using scanning receivers or spectrum analyzers, it is very helpful to provide specific parameters that can be used in comparing the attributes and usefulness of such broadband antennas. Various CISPR publications specify particular antennas to be used in making measurements. Tuned half-wave resonant dipoles are most notably mentioned above 80 MHz. Generally, other types of antennas, normally broadband in nature, can be used provided the results are equivalent to those obtained with the specified antenna. The comparison of these broadband antennas to the specified antennas or to other broadband antennas will be aided by listing appropriate parameters. These parameters shall be specified as part of any CISPR contribution recommending new antenna usage. Antenna manufacturers shall also use this information as guidance in specifying the most useful aspects of broadband antennas used in making interference measurements. It is not the intent of CISPR, however, to show a preference for any particular broadband antenna over that for tuned dipoles.

#### **A.2 Broadband antenna parameters**

Broadband antennas used for CISPR measurements are those antennas that are linearly polarized and are intended for use over a wide frequency range. This does not prevent the use of antennas with limited length adjustment nor the addition of antenna element sections. The impedances of such antennas are typically comprised of both real and imaginary impedances. Other parameters that can be specified are contained below.

##### **A.2.1 Antenna type**

The following parameters describe the physical parameters of broadband antennas that should be provided. Note that some parameters may not apply to each antenna.

##### **A.2.1.1 Antenna style of fixed or variable length or diameter**

If the antenna has a variable length, specify the number of sections that are added or subtracted to change the basic fixed length.

NOTE. Fully tunable antennas are not considered to be broadband and hence would not be specified herein. The diameter of loop antennas are generally not variable.

##### **A.2.1.2 Depth to width ratio or loop diameter**

Provide dimension in metres. For a log periodic array, for example, the length of the boom along the measurement axis and the width of the largest element would be provided.

### **A.2.1.3 Active or passive antenna**

A broadband antenna is considered an active antenna if it contains amplifiers, preamplifiers, and other non-linear active devices which amplify the signal and or shape the frequency response.

### **A.2.1.4 Mounting arrangement**

Provide any special mounting requirements beyond those which can be accommodated by a typical tripod or antenna positioner.

### **A.2.1.5 Connector type**

Specify BNC, N, SMA, etc. as appropriate.

### **A.2.1.6 Balun type**

Specify if balun is discrete, distributed, tunable, etc.

## **A.2.2 Specification of the antenna**

### **A.2.2.1 Frequency range**

Specify the frequency range in megahertz or kilohertz where the antenna operates within its characteristics. If there is a defined fall-off characteristic in decibels per octave at either end of the range, so specify.

### **A.2.2.2 Gain and antenna factor**

#### **A.2.2.2.1 Gain**

Specify typical or actual gain in decibels relative to an isotropic radiator (dBi).

#### **A.2.2.2.2 Antenna factor**

Specify typical or actual antenna factor in decibels per metre.

Both gain and antenna factor should be measured using the calibration procedure in A.2.3.1.

### **A.2.2.3 Directivity and pattern for linearity polarization**

Specify antenna pattern and directivity in degrees with a polar plot in both the E and H planes. For less directional antennas, specify the front-to-back ratio in decibels. If omnidirectional, so state.

### **A.2.2.4 VSWR and impedance**

Indicate the maximum VSWR and nominal input impedance in ohms.

### **A.2.2.5 Active antenna performance**

For antennas with active amplified gain, specify the intermodulation product levels, its electric and magnetic field strength immunity level from outside disturbances, and any appropriate check to determine overload or improper operation.

### **A.2.2.6 Power handling**

For immunity, use specified maximum and transient power handling capability in watts.

### **A.2.2.7 Other conditions**

Specify the temperature and humidity range in which the antenna must operate and any precautions if used in an unprotected area exposed to the weather.

## **A.2.3 Antenna calibration**

### **A.2.3.1 Method of calibration for emission measurements**

Identify the method used for calibration, i.e.:

- a) calculated (indicate formula used);
- b) measured (specify the method or standard used or the traceability to national calibration laboratory, and whether antennas are calibrated individually).

NOTE For immunity measurements, field strength calibrations are generally made using a secondary calibrated antenna located at the place of the appliance being subjected to the radiation. Hence, no calibrations are required on the transmit antenna.

### **A.2.3.2 Frequency interval**

Indicate the frequencies in megahertz or kilohertz used during the calibration process; if a swept frequency procedure is used, so state.

### **A.2.3.3 Accuracy of calibration**

Specify the nominal accuracy of the calibration in  $\pm$  decibels. Indicate the worst case accuracy and the portion of the frequency band where that occurs.

### **A.2.3.4 Correlation with preferred or specified antennas**

If the antenna is to be substituted for a preferred or specified antenna cited in a CISPR publication, indicate all correlation factors in decibels to equate the broadband antenna results to those of the preferred or specified antenna. Also indicate any conversion factor used to convert from the magnetic field intensity or vice versa or for any other conversion to a measurement unit other than a field strength quantity.

### **A.2.3.5 Units**

Specify calibration in units that are necessary to make magnetic or electric field strength emission measurements.

## **A.2.4 Antenna user information**

### **A.2.4.1 Antenna use**

Provide a description of the use of the antenna. Ensure that any special precautions or limitations are cited to reduce the chance of misuse.

### **A.2.4.2 Physical limitations**

Indicate if there are any physical limitations in using the antenna such as the following:

- a) minimum height above the ground plane;
- b) preferred polarization with respect to the ground plane;

- c) special use, i.e. use as a receive antenna or a transmit antenna only. Normally, this is limited to the power handling capability of the balun for passive antennas or the non-bidirectional characteristics for active antennas;
- d) simple ohmic check to determine continuity integrity of antenna;
- e) minimum separation of the closest antenna element to the appliance being measured.

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## Annex B (normative)

### Monopole (1 m rod antenna) performance equations and characterization of the associated antenna matching network <sup>5)</sup>

#### B.1 Description

##### B.1.1 Introduction of the monopole (1 m rod) antenna system

Monopole (rod) antennas are typically used at frequencies below 30 MHz but are sometimes used at higher frequencies. Because of the long wavelength associated with the low frequency range, methods used to calibrate or characterize antennas at higher frequencies are not applicable. The techniques defined in this annex are applicable for frequencies up to 30 MHz. Using due care, this method has been used commercially with small (less than 1 dB) error.

The primary method for traceability of antenna factor to national standards is to illuminate the whole antenna by a plane wave. An alternative method, capacitor substitution of the monopole element, is contained in this annex. Although it is possible to determine the antenna factor by the capacitor substitution method, it requires expert knowledge to achieve the true antenna factor to within  $\pm 1$  dB during the actual calibration process. This is especially the case when designing jigs for types of antenna whose monopole element is not attachable by a coaxial connector. Finally, care in the use of the capacitor substitution method is required especially at frequencies above 10 MHz and for active antennas.

##### B.1.2 Monopole (rod) antenna performance equations

The following equations are used to determine the effective height, self-capacitance and height correction factor of rod or monopole antennas of unusual dimensions.

They are valid only for cylindrical rod antennas shorter than  $\lambda/8$  [8] <sup>6)</sup>.

$$h_e = \frac{\lambda}{2\pi} \tan \frac{\pi h}{\lambda} \quad [1], [2], [3] \quad (\text{B.1})$$

$$C_a = \frac{55,6h}{\left(\ln \frac{2h}{a}\right) - 1} \frac{\tan \frac{2\pi h}{\lambda}}{\frac{2\pi h}{\lambda}} \quad [3], [4], [5], [6], [7], [8] \quad (\text{B.2})$$

$$C_h = 20 \log h_e \quad (\text{B.3})$$

where

$h_e$  is the effective height of the antenna, in metres;

$h$  is the actual height of the rod element, in metres;

$\lambda$  is the wavelength, in metres;

<sup>5)</sup> This annex is based on IEEE 291-1991 (see Clause B.5).

<sup>6)</sup> Figures in square brackets refer to the reference documents cited in Clause B.5.

$C_a$  is the self-capacitance of the rod antenna, in picofarads;

$a$  is the radius of the rod element, in metres;

$C_h$  is the height correction factor, in dB(m).

## B.2 Matching network characterization method

The equivalent capacitance substitution method uses a dummy antenna in place of the actual rod element. The primary component of the dummy antenna is a capacitor equal to the self-capacitance of the rod or monopole. This dummy antenna is fed by a signal source and the output from the matching network or base unit of the antenna is measured using the test configuration shown in Figure B.1. The antenna factor (AF) in dB(1/m) is given by Equation (B.4).

$$AF = V_D - V_L - C_h \quad (\text{B.4})$$

where

$V_D$  is the measured output of the signal generator, in dB( $\mu$ V);

$V_L$  is the measured output of the matching network, in dB( $\mu$ V),

$C_h$  is the height correction factor (for the effective height), in dB(m).

For the monopole (1 m rod) antenna commonly used in EMC measurements, the effective height ( $h_e$ ) is 0,5 m, the height correction factor ( $C_h$ ) is –6 dB(m) and the self-capacitance ( $C_a$ ) is 10 pF.

NOTE See B.1.2 to calculate the effective height, height correction factor and self-capacitance of rod antennas of unusual dimensions.

Either of two procedures shall be used: the method of B.2.1, the network analyser, or the method of B.2.2, the signal generator and radio-noise meter method. The same dummy antenna is used in both procedures. See Clause B.3 for guidance in making a dummy antenna. Measurements shall be made at a sufficient number of frequencies to obtain a smooth curve of antenna factor versus frequency over the operating range of the antenna, or 9 kHz to 30 MHz, whichever is smaller.

### B.2.1 Network analyser procedure

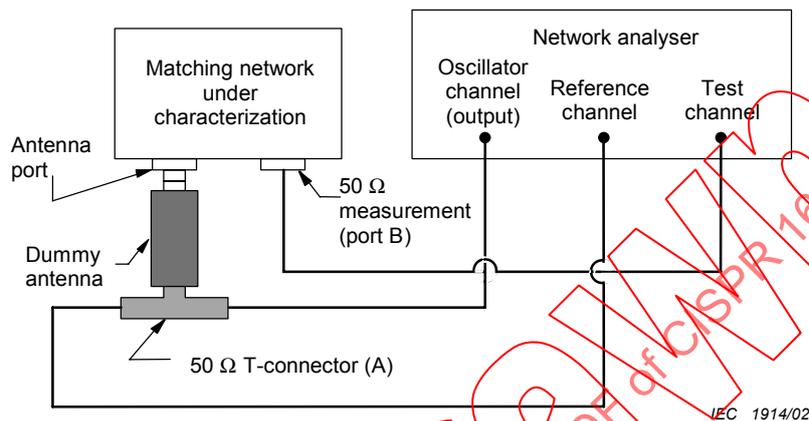
- Calibrate the network analyser with the cables to be used in the measurements.
- Set up the matching network to be characterized and the measuring equipment as shown in Figure B.1.
- Subtract the signal level (in dB( $\mu$ V)) in the test channel from the signal level (in dB( $\mu$ V)) in the reference channel and subtract  $C_h$  (–6 dB for the 1 m rod) to obtain the antenna factor (in dB(1/m)) of the antenna.

NOTE Attenuator pads are not needed with the network analyser because the impedances of the channels in the network analyser are very nearly 50  $\Omega$  and any errors are corrected during network analyser calibration. Attenuator pads may be used, if desired, but including them complicates the network analyser calibration.

### B.2.2 Radio-noise meter and signal generator procedure

- Set up the matching network to be characterized and the measuring equipment as shown in Figure B.2.
- With the equipment connected as shown and a 50  $\Omega$  termination on the T-connector (A), measure the received signal voltage  $V_L$  (in dB( $\mu$ V)) at the RF port (B).

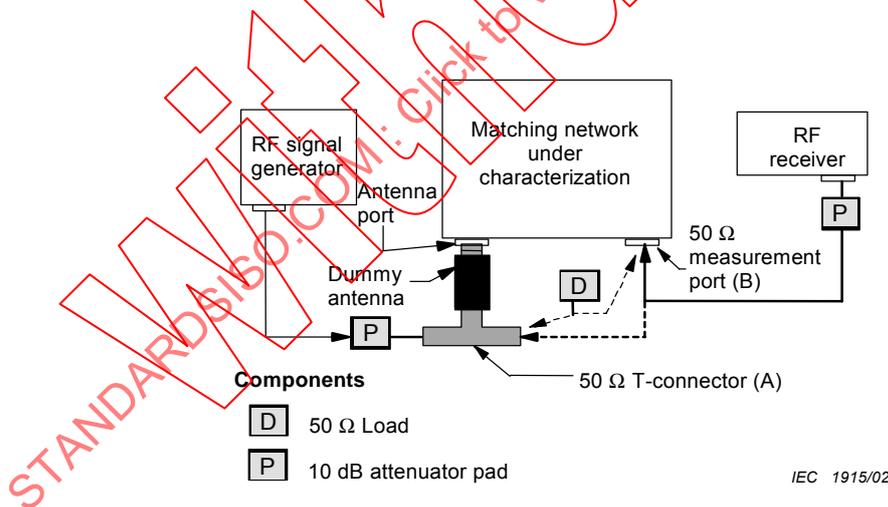
- c) Leaving the RF output of the signal generator unchanged, transfer the 50 Ω termination to the RF port (B) and transfer the receiver input cable to the T-connector (A). Measure the drive signal voltage  $V_D$  (in dB(μV)).
- d) Subtract  $V_L$  from  $V_D$  and subtract  $C_h$  (-6 dB for the 1 m rod) to obtain the antenna factor (in dB(1/m)) of the antenna.



NOTE 1 Place the dummy antenna as close to the EUT port as possible. Place the T-connector as close to the dummy antenna as possible. Use the same length and type of cables between the T-connector and the reference channel input, and the T-connector and the 50 Ω measuring port test channel.

NOTE 2 Attenuator pads are not needed with the network analyser and are not recommended.

Figure B.1 – Method using network analyser



NOTE 1 Place the dummy antenna as close to the EUT port as possible. Place the T-connector as close to the dummy antenna as possible.

NOTE 2 If the VSWR of receiver and signal generator is low, pads may not be needed or may be reduced to 6 dB or 3 dB.

NOTE 3 The dummy antenna may incorporate other matching components to control VSWR at its input and signal generator level at measuring ports.

Figure B.2 – Method using radio-noise meter and signal generator

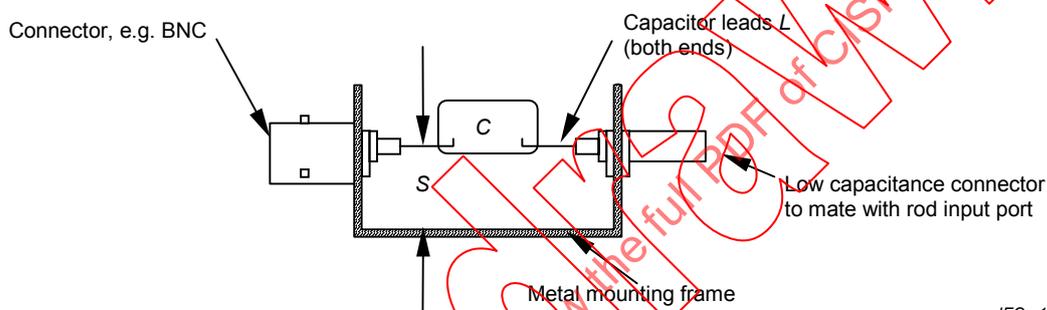
The 50  $\Omega$  termination shall have a very low standing-wave ratio (SWR) (less than 1,05:1). The radio-noise meter shall be calibrated and have a low SWR (less than 2:1). The output of the signal generator shall be frequency and amplitude stable.

NOTE The signal generator need not be calibrated, since it is used as a transfer standard.

### B.3 Dummy antenna considerations

The capacitor used as the dummy antenna shall be mounted in a small metal box or on a small metal frame. The leads shall be kept as short as possible, but no longer than 8 mm, and spaced 5 mm to 10 mm from the surface of the metal box or frame. See Figure B.3.

The T-connector used in the antenna factor measurement set-up may be built into the dummy antenna box. The resistor pad to provide impedance matching to the generator may also be built into the dummy antenna box.



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#### Components

- C antenna capacitance ( $C_a$ ) calculated from Equation (B.2), 5 % tolerance, silver mica.
- S lead spacing, 5 mm to 10 mm (10 mm from all surfaces if enclosed in a box).
- L lead length, as short as possible but not greater than 8 mm (total lead length not greater than 40 mm, including both capacitor leads and length of rod port connector).

Figure B.3 – Example of mounting capacitor in dummy antenna

### B.4 Application of the monopole (rod) antenna

A monopole rod antenna is typically designed to be used with a counterpoise or to be mounted on a groundplane. To obtain correct field strength values, the manufacturer's instructions or recommendations regarding the use of the counterpoise or groundplane should be followed.

If the antenna uses a telescoping rod element, the element shall be extended to the length specified in the manufacturer's instruction.

Many measurement standards specify that the counterpoise of a monopole (rod) antenna shall be bonded to the groundplane or test bench groundplane. The requirements of the measurement standard shall be met.

## B.5 Reference documents

- [1] IEEE 291-1991, *IEEE Standard Methods for Measuring Electromagnetic Field Strength of Sinusoidal Continuous Waves, 30 Hz to 30 GHz*. IEEE, Inc., 445 Hoes Lane, PO Box 1331, Piscataway, NJ 08855-1331 USA, p. 28-29.
- [2] GREENE, FM. NBS Field-Strength Standards and Measurements (30 Hz to 1000 MHz). *Proc. IEEE*, No. 6, June 1967, vol. 55, p. 974-981.
- [3] SCHELKUNOFF, SA. and FRIIS, HT. *Antennas: Theory and Practice*. New York: John Wiley and Sons, Inc., 1952, p. 302-331.
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- [6] HALLÉN, E. Theoretical Investigation into the Transmitting and Receiving Qualities of Antennas. *Nova Acta Soc. Sci. Upsaliensis*, Ser. IV, 11, No. 4, 1938, p. 1-44.
- [7] KING, RWP., *Theory of Linear Antennas*, Harvard University Press, Cambridge, MA, 1956, p.16-17, 71, 184 and 487.
- [8] *The Radio Frequency Interference Meter NAVSHIPS 94810*, by The Staff of the Moore School of Electrical Engineering, University of Pennsylvania, 1962, p. 36-38.

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## Annex C (normative)

### Loop antenna system for magnetic field induced current measurements in the frequency range of 9 kHz to 30 MHz

#### C.1 Introduction

This annex sets forth information and data concerning the loop antenna system (LAS) to measure the current induced in the LAS by the magnetic field emitted by a single EUT, positioned in the centre of the LAS, in the frequency range of 9 kHz to 30 MHz. Subclause 4.7 of this publication and CISPR 16-2-3 refer to this LAS.

A description of the LAS is given, as well as the method of validation of the antennas of the LAS. Conversion factors are given to relate magnetic field induced current data to magnetic field data which would have been obtained when the same EUT was measured using a single-loop magnetic field antenna positioned at a specified distance from that EUT.

#### C.2 Construction of the loop antenna system (LAS)

The LAS (see Figure C.1) consists of three mutually perpendicular large-loop antennas (LLAs), described in Clause C.3. The entire LAS is supported by a non-metallic base.

A 50  $\Omega$  coaxial cable between the current probe of an LLA and the coaxial switch, and between this switch and the measuring equipment, shall have a surface transfer impedance smaller than 10 m $\Omega$ /m at 100 kHz and 1 m $\Omega$ /m at 10 MHz. This requirement is met when using, for example, double-braided shield RG 223/U coaxial cable.

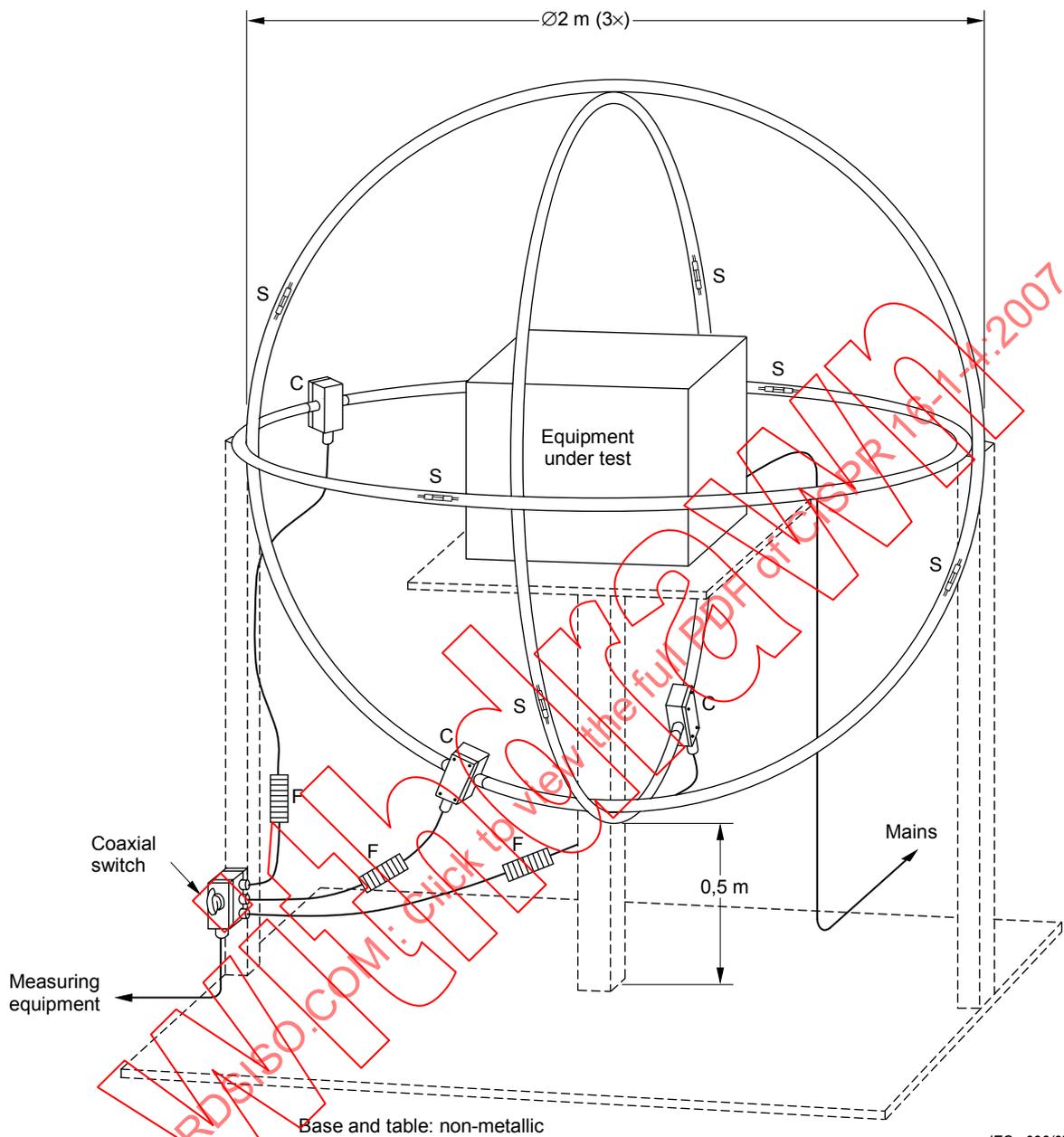
All connectors shall have a surface transfer impedance comparable with that of the coaxial cable. This requirement is met, for example, when using good quality BNC collet-lock type connectors (see IEC 60169-8\*).

All cables shall be equipped with ferrite absorbers, F in Figure C.1, providing a common-mode series resistance of  $R_s > 100 \Omega$  at 10 MHz. This requirement is met when constructing the ferrite toroid from, for example, 12 rings of type 3E1 from Ferroxcube (minimum size in millimetres: 29 O.D.  $\times$  19 I.D.  $\times$  7,5 Ht).

#### C.3 Construction of a large-loop antenna (LLA)

A large-loop antenna (LLA) of the LAS is constructed from coaxial cable of which the surface transfer impedance has been specified in Clause C.2. In addition, the resistance of the inner conductor of the LLA shall be sufficiently low (see Note 1). Both requirements are met, for example, when using double-braided shield RG 223/U coaxial cable.

\* IEC 60169-8:1978, *Radio-frequency connectors – Part 8: RF coaxial connectors with inner diameter of outer conductor 6,5 mm (0,256 in) with bayonet lock – Characteristic impedance 50 ohms (Type BNC)*



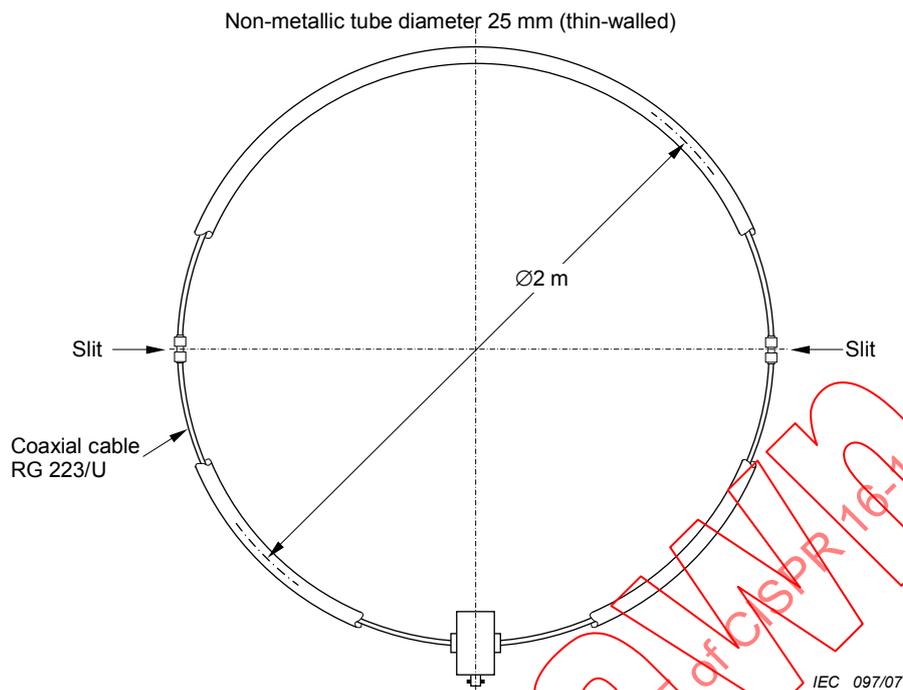
IEC 096/07

**Components**

- S: antenna slit
- C: current probe
- F: ferrite absorber

**Figure C.1 – The loop-antenna system, consisting of three mutually perpendicular large-loop antennas**

To keep the loop in its circular shape and to protect the slit construction, as in the example of Figure C.2, the cable is inserted in a thin walled non-metallic tube with inner diameter of approximately 25 mm. Other non-metallic constructions serving the same purposes may be used.

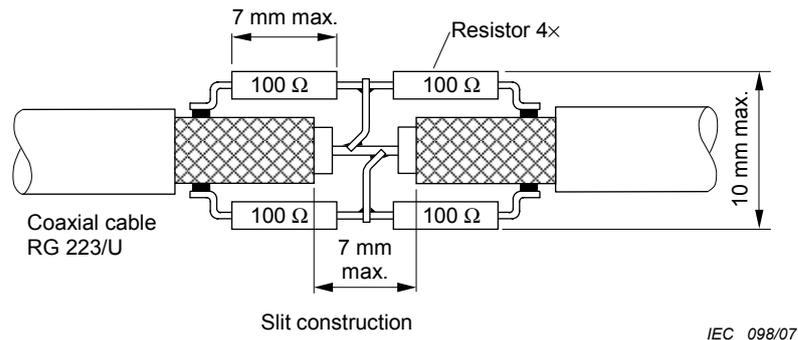


**Figure C.2 – A large-loop antenna containing two opposite slits, positioned symmetrically with respect to the current probe C**

The loop diameter has been standardized to be  $D = 2$  m. If necessary, e.g. the case of large EUT,  $D$  may be increased. However, in the frequency range up to 30 MHz, the maximum allowable diameter is 4 m. Further increase of the diameter would result in non-reproducible resonances of the LAS response at the high-frequency end of the measuring range.

It should be noted that by increasing the diameter, its sensitivity to ambient noise increases proportionally to the diameter, and its sensitivity to wanted signals is inversely proportional with the diameter squared.

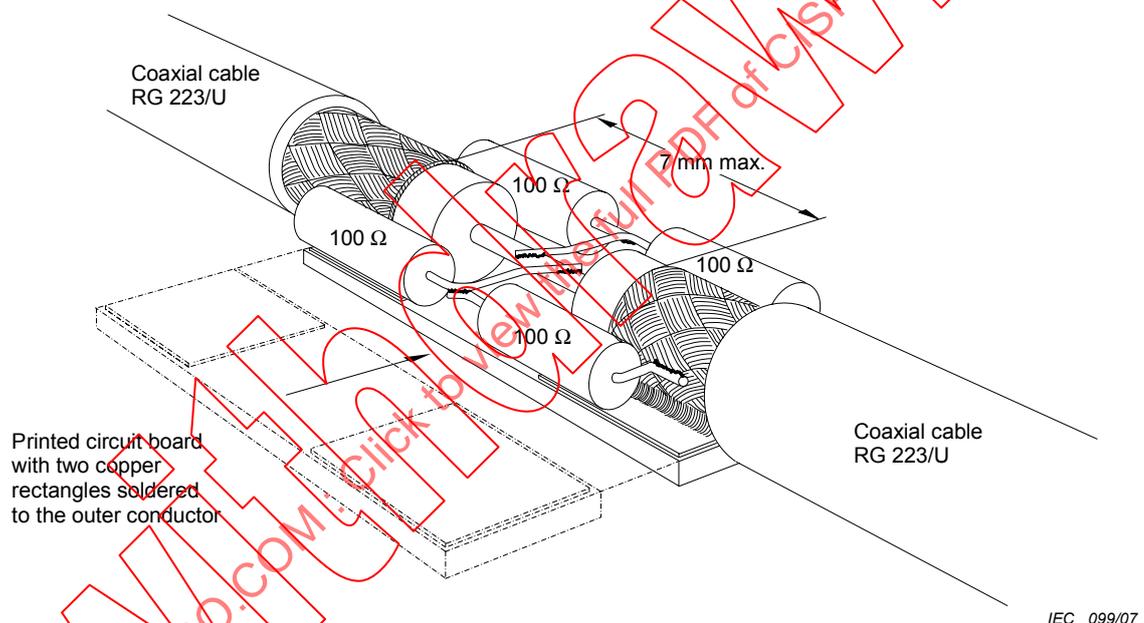
An LLA contains two opposite slits, positioned symmetrically with respect to the current probe of the LLA (see Figure C.2). Such a slit, made in the outer conductor of the coaxial antenna cable as shown in Figure C.3, shall have a width of less than 7 mm. The slit is bridged by two parallel sets of 100  $\Omega$  resistors in series. The centre of each series circuit is connected to the inner conductor of the coaxial antenna cable.



IEC 098/07

**Figure C.3 – Construction of the antenna slit**

At each side of the slit, the outer conductor of the coaxial antenna cable may be bonded to a strap of printed circuit board material with two copper rectangles, separated by at least 5 mm, in order to obtain a rigid slit construction (see Figure C.4).



IEC 099/07

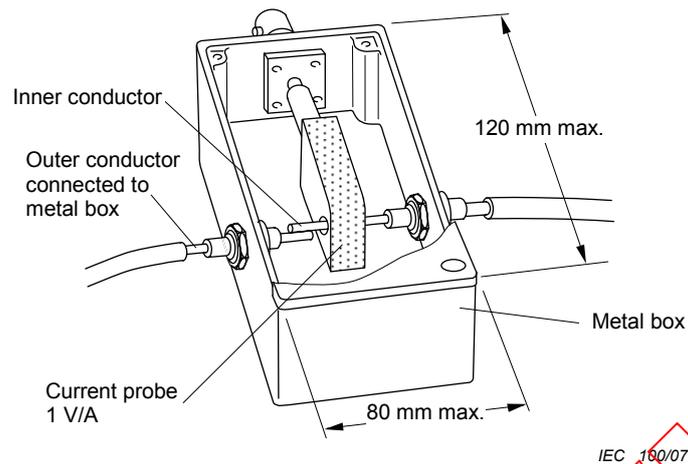
**Figure C.4 – Example of antenna-slit construction using a strap of printed circuit board to obtain a rigid construction**

The current probe around the inner conductor of the coaxial antenna-cable shall have a sensitivity of 1 V/A over the frequency range of 9 kHz to 30 MHz. The insertion loss of the current probe shall be sufficiently low (see Note 1).

The outer conductor of that cable shall be bonded to the metal box containing the current probe (see Figure C.5). The maximum dimensions of this box are the following: width 80 mm, length 120 mm and height 80 mm.

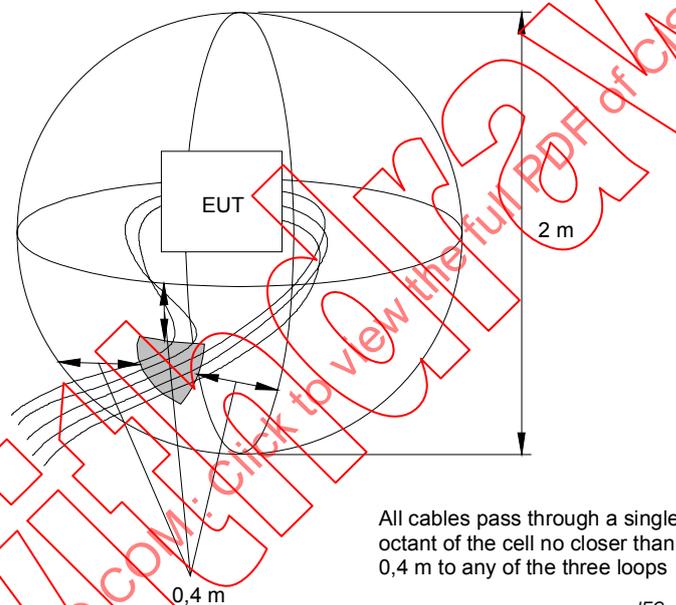
NOTE 1 To obtain a flat frequency response of the LLA at the lower end of the frequency range of 9 kHz to 30 MHz, the insertion loss  $R_c$  of the current probe should be much smaller than  $2 \pi f L_c$  at  $f = 9$  kHz, where  $L_c$  represents the inductance of the current probe. In addition,  $(R_c + R_i) \ll X_i = 2 \pi f L$  at 9 kHz, where  $R_i$  is the resistance of the inner conductor of the loop and  $L$  is the loop inductance. This inductance is about 1,5  $\mu$ H/m of circumference, hence for the standardized LLA,  $X_i \approx 0,5 \Omega$  at  $f = 9$  kHz.

NOTE 2 To avoid unwanted capacitive coupling between the EUT and the LAS, the distance between the EUT and components of the LLA should be at least 0,10 times the loop diameter. Particular attention must be paid to the leads of an EUT. Cables should be routed together and leave the loop volume in the same octant of the cell, no closer than 0,4 m to any of the LAS loops (see Figure C.6).



IEC 180/07

**Figure C.5 – Construction for the metal box containing the current probe**



All cables pass through a single octant of the cell no closer than 0,4 m to any of the three loops

IEC 1 009/97

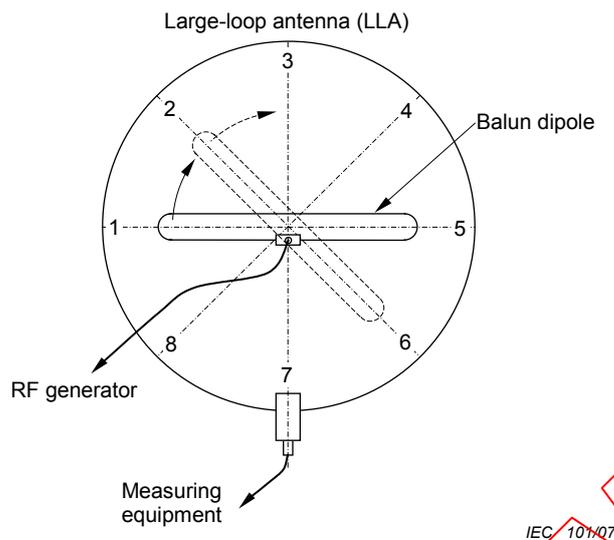
**Figure C.6 – Example showing the routing of several cables from an EUT to ensure that there is no capacitive coupling from the leads to the loop**

#### C.4 Validation of a large-loop antenna (LLA)

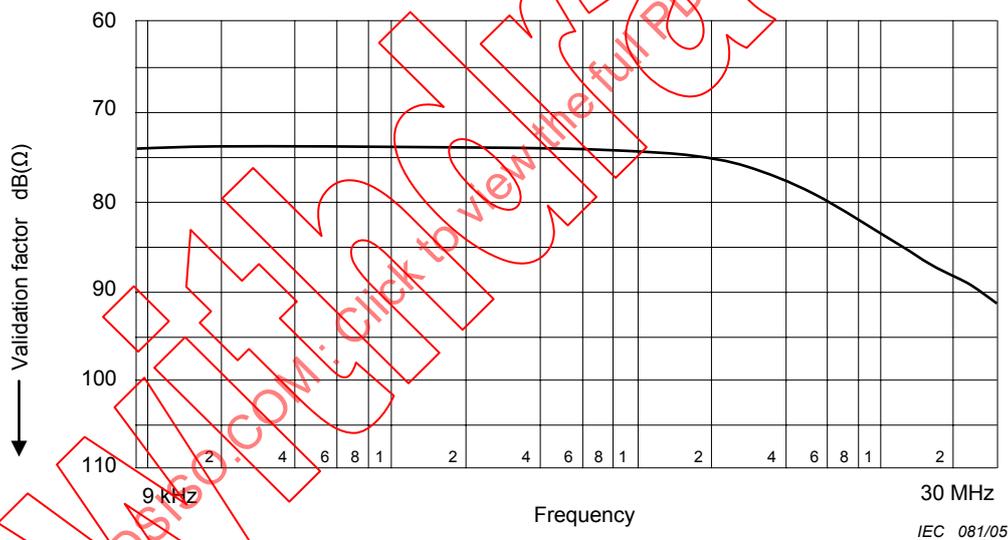
The validation and calibration of a large-loop antenna (LLA) of the loop antenna system is carried out by measuring the current induced in the LLA by the balun-dipole connected to a  $50 \Omega$  RF generator, described in Clause C.5. The magnetic field emitted by that dipole allows verification of the magnetic field sensitivity of the LLA. The electric field emitted by the balun-dipole shows that the electric field sensitivity of the LLA is sufficiently low.

The induced current shall be measured as a function of frequency in the range of 9 kHz to 30 MHz at the 8 positions of the balun-dipole in Figure C.7. During this measurement, the balun dipole is in the plane of the LLA under test.

In each of the eight positions, the validation factor [expressed in  $\text{dB}(\Omega) = 20 \log(V_{g0}/I_l)$ ] of the open circuit voltage of the RF generator ( $V_{g0}$ ) and the measured current ( $I_l$ ) shall not deviate more than  $\pm 2$  dB from the validation factor given in Figure C.8.



**Figure C.7 – The eight positions of the balun-dipole during validation of the large-loop antenna**



**Figure C.8 – Validation factor for a large loop-antenna of 2 m diameter**

The validation factor given in Figure C.8 is valid for a circular LLA with a standardized diameter  $D = 2$  m. If the diameter of a circular LLA differs from  $D = 2$  m, the validation factor for the non-standardized LLA can be derived from the data given in Figures C.8 and C.11 (Clause C.6).

### C.5 Construction of the balun-dipole

The balun-dipole, Figure C.9, has been designed to emit simultaneously a magnetic field, which should be measured by the LLA, and an electric field, which should be rejected by the LLA.

The balun-dipole is constructed from RG 223/U coaxial cable. It has a width  $W = 150$  cm and a height  $H = 10$  cm (cable centre to cable centre distances), as depicted in Figure C.9.

A slit in the outer conductor of the coaxial cable divides the dipole in two halves. One half of this dipole, the right-hand half in Figure C.9, is short-circuited near the slit as well as near the connector. Short-circuited means that the inner and outer conductors of the coaxial cable are electrically bonded together. This half is connected to the reference-ground of the BNC connector. The inner conductor of the coaxial cable, forming the left-hand half of the dipole in Figure C.9, is connected to the centre-pin of the BNC connector and its outer conductor to the reference ground of that BNC connector.

A small metal box is used to screen the connections near the dipole connector. The outer conductor of the two halves of the coaxial dipole cable are bonded to this box, as is the reference ground of the BNC connector.

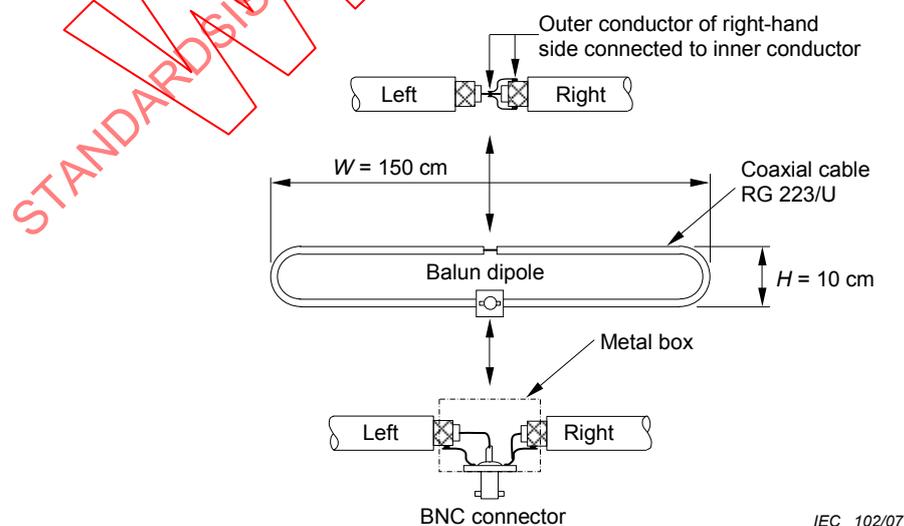
To obtain a rigid construction, the dipole is supported by a non-conductive base.

### C.6 Conversion factors

This clause deals with the factor which converts the current ( $I$ ) induced in the LLA by the EUT into a magnetic field strength  $H$  at a specified distance from the EUT (see Figure C.10). It also deals with the factor which converts the current measured in an LLA with a non-standardized diameter to a current which would have been measured using an LLA with the standardized diameter of  $D = 2$  m (see Figure C.11).

The conversion factor in Figure C.10 applies to a source of magnetic field positioned in the centre of the LLA with its dipole moment perpendicular to the plane of that LLA. It should be noted that with the loop antennas specified in 4.2, the loop antenna is always positioned in a vertical plane and the EUT is only rotated around its vertical axis. Hence, in that case only the horizontal dipole moments, i.e. the dipole moments parallel to the ground plane, are measured. Consequently, in the case of a vertical dipole moment the conversion factor cannot be used to compare results of both measuring methods. However, the factor can be used when in the magnetic field measuring method the loop antenna would be positioned in a horizontal plane, or when in that method the EUT would be tilted through  $90^\circ$ , so that the relevant vertical dipole moment is changed into a horizontal one.

If the actual position of a disturbance source inside an EUT is at a distance less than 0,5 m from the centre of the standardized LAS, the measuring results differ by less than 3 dB from those with that source in the centre.



IEC 102/07

Figure C.9 – Construction of the balun-dipole